



Design Example Report

Title	2.7 W StackFET Buck Power Supply Using LinkSwitch™-TN2 LNK3206GQ
Specification	60 VDC – 950 VDC Input (Withstand 1100 VDC); 18.0 V / 150 mA Output
Application	Automotive Application
Author	Applications Engineering Department
Document Number	DER-719Q
Date	July 23, 2021
Revision	1.1

Summary and Features

- Automotive graded BOM
- 750 V power MOSFET rating
- Highly integrated solution
- Lowest possible component count
- No optocoupler required for regulation
- 950 VDC input. Can withstand 1100 VDC input
- <±5% load regulation
- 105 °C ambient temperature operation

PATENT INFORMATION

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Important Note: Although this board is designed to satisfy safety isolation requirements, the engineering prototype has not been agency approved.



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1 Introduction

This engineering report describes a StackFET buck converter designed to provide a non-isolated nominal output voltage of 18 V at 150 mA load from a wide input voltage range of 60 VDC to 950 VDC and can withstand 1100 VDC for automotive application. This power supply utilizes the LNK3206GQ from the LinkSwitch-TN2 family for Automotive of ICs.

This document contains the complete power supply specifications, bill of materials, transformer construction, circuit schematic and printed circuit board layout, along with performance data and electrical waveforms.

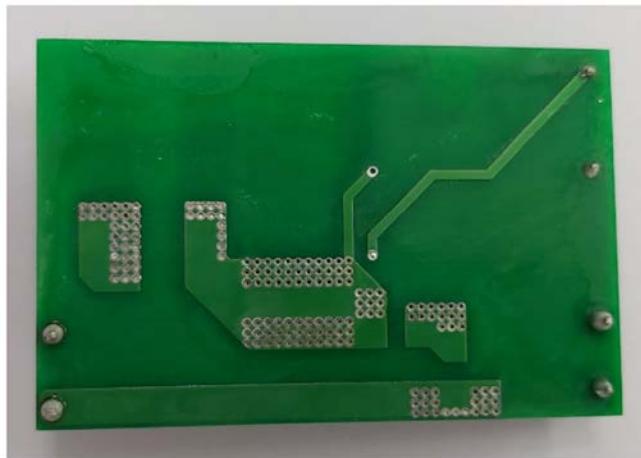
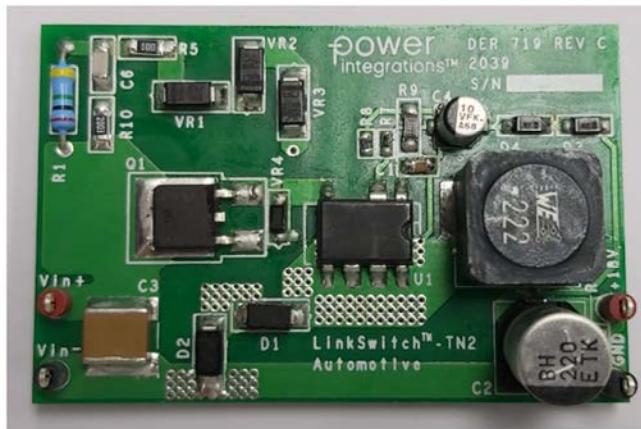


Figure 1 – Populated Circuit Board.



2 Power Supply Specification

The table below represents the minimum acceptable performance of the design. Actual performance is listed in the results section.

Description	Symbol	Min	Typ	Max	Units	Comment
Input Voltage	V_{IN}	60		950	VDC	Withstand 1100 VDC Input.
Output1 Output Voltage No-Load Output Voltage Output Ripple Voltage Output Current	V_{OUT} V_{OUT} V_{RIPPLE} I_{OUT}	17.1	18	18.9 19.08 180 150	V V mV mA	$\pm 5\%$ Regulation at No-Load. 20 MHz Bandwidth.
Total Output Power Continuous Output Power	P_{OUT}			2.7	W	
Efficiency Full Load Efficiency @ 60 V_{IN} Full Load Efficiency @ 360 V_{IN} Full Load Efficiency @ 950 V_{IN}	η η η	73			%	Measured at P_{OUT} 25 °C. Measured at P_{OUT} 25 °C. Measured at P_{OUT} 25 °C.
Environmental Ambient Temperature	T_{AMB}	0		105	°C	Free Convection, Sea Level.



3 Schematic

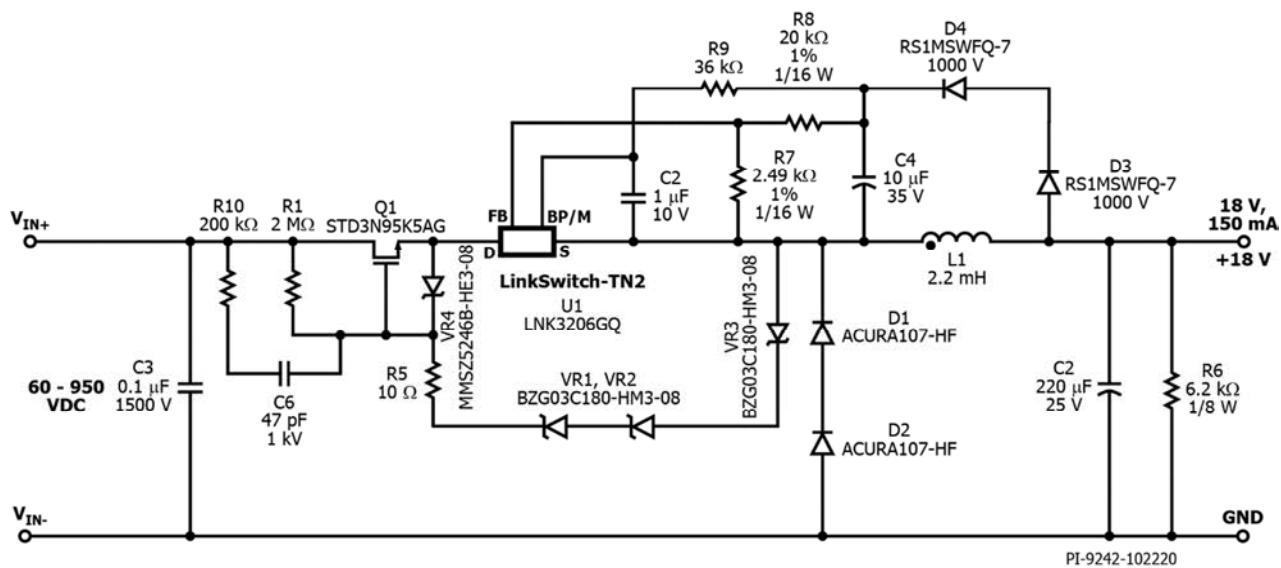


Figure 2 – Schematic.

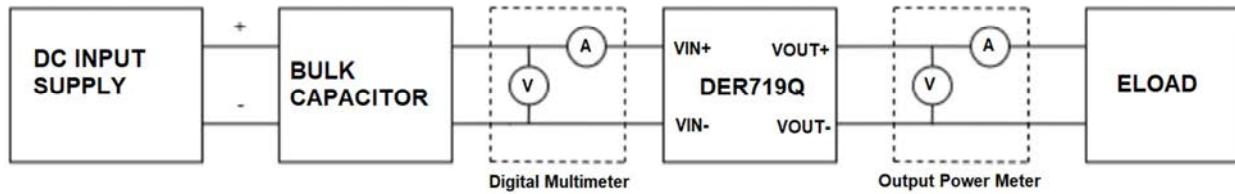


Figure 3 – Test Set-up.



4 Circuit Description

The schematic in Figure 2 shows a StackFET buck converter using LNK3206GQ. The circuit provides a non-isolated 18 V, 150 mA continuous output. This configuration enhances its withstand voltage so that the circuit can be applied with higher input voltage.

In this design, automotive battery will be the intended input source of the buck circuit. To simulate battery stiff voltage source, bulk capacitors were added in the test set-up during the test as shown in Figure 3.

All components selected in this report are automotive compliant.

4.1 *Input Filter*

A bypass capacitor C1 is added to provide a local instantaneous charge and a stable DC bus to the buck converter.

4.2 *Power Stage*

MOSFET Q1 and LNK3206GQ are arranged in the StackFET configuration. The resistor R1 performs the function of initially turning on the MOSFET, so that the BYPASS pin capacitor can be charged from the input voltage. Resistor R10 and C6 are used to balance the voltage sharing between MOSFET and LNK3206GQ during start-up.

The Zener diodes (VR1, VR2, and VR3) performs dual function in the circuit. It limits the voltage across LNK3206GQ to less than 600 V. It also provides gate current to Q1 (from the charge stored across its junction capacitance) while turning it on and off. Series resistor R5 limits gate charging current so that any tendency of high frequency oscillation can be damped out. Zener VR4 limits the gate to source voltage of Q1.

During normal operation the device is powered from output via a current limiting resistor R9.

The operation of U1 is unaffected by the StackFET configuration. When LNK3206GQ turns on Q1 also turn on, applying the input voltage across the output inductor L1. Once the primary current reaches the internal current limit of LNK3206GQ IC, the MOSFET turns-off. The on-time for each switching cycle is set by the inductance value of L1, LinkSwitch-TN2 auto current limit and the high-voltage DC input bus across C3. Output regulation is accomplished by skipping switching cycles in response to an ON/OFF feedback signal applied to the FEEDBACK (FB) pin. This differs significantly from traditional PWM schemes that control the duty factor (duty cycle) of each switching cycle. Unlike TinySwitch, the logic of the FB pin has been inverted in LinkSwitch-TN. This allows a very simple feedback scheme to be used when the device is used in the buck converter configuration. Current into the FB pin greater than 49 μ A will inhibit the switching of the internal MOSFET, while current below this allows switching cycles to occur.



4.3 ***Output Rectification***

During the ON time of U1, current ramps in L1 and is simultaneously delivered to the load. During the OFF time, the inductor current ramps down via free-wheeling diode D1 and D2 into C2 and is delivered to the load. Diode D1 and D2 should be selected as an ultrafast diode (low T_{RR}) due to high ambient temperature operation. To meet 80% diode voltage derating requirement, two freewheeling diode in series was implemented.

Capacitor C2 should be selected to have an adequate ripple current rating (very low ESR type). Please see the spreadsheet output capacitor section.

4.4 ***Output Feedback***

The voltage across L1 is rectified and smoothed by D3, D4, and C4 during the off-time of U1. To provide a feedback signal, the voltage developed across C4 is divided by R7 and R8 and connected to U1's FB pin. The values of R7 and R8 are selected such that at the nominal output voltage, the voltage on the FB pin is 2 V. Resistor R7 and R8 can be optimized for better output voltage regulation and efficiency. This voltage is specified for U1 at an FB pin current of 49 μ A. This allows simple feedback to meet the required overall output tolerance of $\pm 5\%$ at rated output current.

4.5 ***PCB***

Printed Circuit Board (PCB) should be rigid enough to survive the harsh environment of automotive application. FR-4 High Tg PCB material was used in the design.



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5 PCB Layout

PCB copper thickness is 2oz. PCB thickness is 1 mm. PCB Material FR4 High Tg 170.

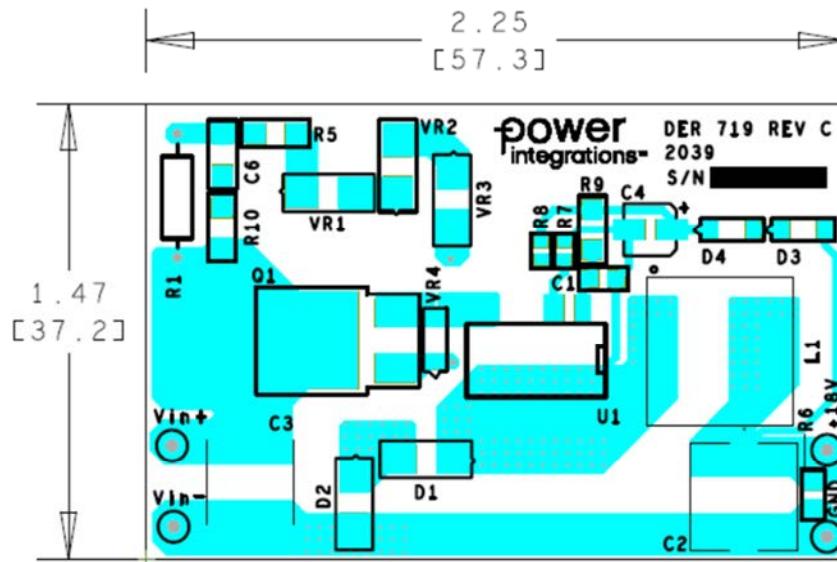


Figure 4 – Printed Circuit Board, Top View.

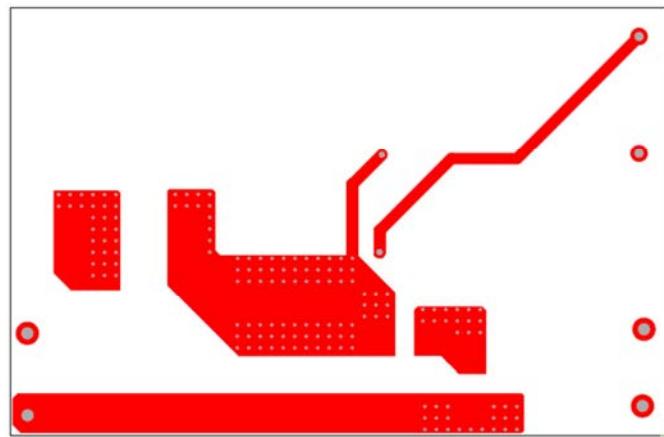


Figure 5 – Printed Circuit Board, Bottom View.



6 Bill of Materials

Item	Qty	Ref Des	Description	Mfg Part Number	Mfg
1	1	C1	1 μ F \pm 10% 10 V Ceramic X7R 0805	C0805C105K8RACAUTO	Kemet
2	1	C2	220 μ F, 20%, 25 V, Electrolytic, (8.30 mm x 8.30 mm), SMD	25TKV220M8X10.5	Rubycon
3	1	C3	0.1 μ F \pm 10% 1500 V (1.5kV) Ceramic X7R 2225	C2225C104KFRACAUTO	Kemet
4	1	C4	10 μ F, 20%, 35V, Electrolytic, (4.30 mm x 4.30 mm), SMD	EEE-FK1V100UR	Panasonic
5	1	C6	47 pF, 10000 pF, \pm 5%, 1000 V, Ceramic, COG_NP0, 1206	C1206C470JDGACAUTO	KEMET
6	1	D1	Diode, Ultrafast Recovery, 1000 V, 1 A, Surface Mount, DO-214AC (SMA)	ACURA107-HF	Comchip
7	1	D2	Diode, Ultrafast Recovery, 1000 V, 1 A, Surface Mount, DO-214AC (SMA)	ACURA107-HF	Comchip
8	1	D3	Diode Standard, Fast Recovery, 1000 V 1 A SMT SOD-123F	RS1MSWFQ-7	Diodes, Inc.
9	1	D4	Diode Standard, Fast Recovery, 1000 V 1 A SMT SOD-123F	RS1MSWFQ-7	Diodes, Inc.
10	1	L1	2.2 mH Shielded Wirewound Inductor 530 mA 3.75 Ω Max Nonstandard (12 x 12 mm)	7447709222	Wurth
11	1	Q1	N-Channel C 2A (T _c) 45 W (T _c) SMT, DPAK, TO-252-3, DPak (2 Leads + Tab), SC-63	STD3N95K5AG	ST Micro
12	1	R1	RES, 2 M Ω , \pm 5%, 1/4 W Through Hole, Axial, High-Voltage, Pulse Withstanding Metal Film	VR25000002004JR500	Vishay
13	1	R5	RES, 10 Ω , 5%, 1/4 W, Thick Film, 1206	ERJ-8GEYJ100V	Panasonic
14	1	R6	RES, 6.2 k Ω , 5%, 1/8 W, Thick Film, 0805	ERJ-6GEYJ622V	Panasonic
15	1	R7	RES, 2.49 k Ω , 1%, 1/16 W, Thick Film, 0603	ERJ-3EKF2491V	Panasonic
16	1	R8	RES, 20 k Ω , 1%, 1/16 W, Thick Film, 0603	ERJ-3EKF2002V	Panasonic
17	1	R9	RES, 36 k Ω , 5%, 1/4 W, Thick Film, 1206	ERJ-8GEYJ363V	Panasonic
18	1	R10	RES, 200 k Ω , 5%, 1/4 W, Thick Film, 1206	ERJ-8GEYJ204V	Panasonic
19	1	U1	LinkSwitch-TN2, SMD-8C	LNK3206GQ	Power Integrations
20	1	VR1	Zener Diode 180 V 1.25 W \pm 6.39% SMT DO-214AC (SMA)	BZG03C180-HM3-08	Vishay
21	1	VR2	Zener Diode 180 V 1.25 W \pm 6.39% SMT DO-214AC (SMA)	BZG03C180-HM3-08	Vishay
22	1	VR3	Zener Diode 180 V 1.25 W \pm 6.39% SMT DO-214AC (SMA)	BZG03C180-HM3-08	Vishay
23	1	VR4	Zener Diode 16 V 500 mW \pm 5% SMT SOD-123	MMSZ5246B-HE3-08	Vishay

Miscellaneous

Item	Qty	Ref Des	Description	Mfg Part Number	Mfg
1	1	+18V	Test Point, RED, Miniature THRU-HOLE MOUNT	5000	Keystone
2	1	GND	Test Point, BLK, Miniature THRU-HOLE MOUNT	5001	Keystone
3	1	VIN+	Test Point, RED, Miniature THRU-HOLE MOUNT	5000	Keystone
4	1	VIN-	Test Point, BLK, Miniature THRU-HOLE MOUNT	5001	Keystone



7 Design Spreadsheet

DCDC_LinkSwitchTN2-Automotive-Buck_080320; Rev.1.1; Copyright Power Integrations 2020					
	INPUT	INFO	OUTPUT	UNIT	DCDC LinkSwitchTN2-Automotive Buck
ENTER APPLICATION VARIABLES					
VDCMIN	60.00		60.00	V	Minimum DC voltage
VDCMAX	950.00	Info	950.00	V	Entered maximum DC voltage exceeds device datasheet recommendation. Decrease the value to less than or equal to 550V
VOUT	18.00		18.00	V	Output voltage
IOUT	0.150		0.150	A	Average output current
EFFICIENCY_ESTIMATED			0.80		Efficiency estimate at output terminals
EFFICIENCY_CALCULATED			0.78		Calculated efficiency based on real components and operating point
POUT			2.70	W	Continuous output power
INPUT STAGE RESISTANCE			10	Ohms	Input stage resistance in ohms (includes thermistor, filtering components, etc)
PLOSS_INPUTSTAGE			0.032	W	Maximum input stage loss
ENTER LINKSWITCH-TN2-AUTOMOTIVE VARIABLES					
OPERATION MODE			MDCM		Mostly discontinuous mode of operation
CURRENT LIMIT MODE	RED		RED		Choose 'RED' for reduced current limit or 'STD' for standard current limit
PACKAGE			SMD-8C		Select the device package
DEVICE SERIES			LNK3206		Generic LinkSwitch-TN2 device
DEVICE CODE			LNK3206GQ		Required LinkSwitch-TN2 device
ILIMITMIN			0.325	A	Minimum current limit of the device
ILIMITTYP			0.370	A	Typical current limit of the device
ILIMITMAX			0.415	A	Maximum current limit of the device
RDSON			15.50	ohms	Primary switch on-time drain to source resistance at 125degC
FSMIN			62000	Hz	Minimum switching frequency
FSTYP			66000	Hz	Typical switching frequency
FSMAX			70000	Hz	Maximum switching frequency
BVDSS			750	V	Device breakdown voltage
PRIMARY SWITCH PARAMETERS					
VDSON			2.00	V	Primary switch on-time drain to source voltage estimate
VDSOFF		Warning	998	V	Drain voltage stress exceeds breakdown voltage of device. Decrease maximum DC input voltage
DUTY			0.325		Maximum duty cycle
TIME_ON_MIN			0.821	us	Primary switch minimum on-time
IPED_PRIMARYSWITCH			0.000	A	Maximum primary switch pedestal current
IRMS_PRIMARYSWITCH			0.116	A	Maximum primary switch RMS current
PLOSS_PRIMARYSWITCH			0.273	W	Maximum primary switch loss
BUCK INDUCTOR PARAMETERS					
INDUCTANCE_MIN			1980	uH	Minimum design inductance required for current delivery. Note that the chosen inductor must be AEC-Q200 compliant
INDUCTANCE_TYP	2200		2200	uH	Typical design inductance required for current delivery. Note that the chosen inductor must be AEC-Q200 compliant
INDUCTANCE_MAX			2420	uH	Maximum design inductance required for current delivery. Note that the chosen inductor must be AEC-Q200 compliant
TOLERANCE_INDUCTANCE	10		10	%	Tolerance of the design inductance
DC RESISTANCE OF INDUCTOR			2.0	ohms	DC resistance of the buck inductor
FACTOR_KLOSS			0.50		Factor that accounts for "off-state" power loss to be supplied by inductor (usually between 50% to 66%)



IRMS_INDUCTOR			0.222	A	Maximum inductor RMS current
PLOSS_INDUCTOR			0.098	W	Maximum inductor losses
FREEWHEELING DIODE PARAMETERS					
VF_FREEWHEELING			2.50	V	Forward voltage drop across the two freewheeling diodes in series
PIV_RATING			600.00	V	Peak inverse voltage rating of each freewheeling diode
TRR			25	ns	Reverse recovery time of each freewheeling diode
PIV_CALCULATED		Info	998	V	Combined PIV ratings of two diodes should meet 80% voltage derating requirement
IRMS_DIODE			0.219	A	Maximum diode RMS current
PLOSS_DIODE			0.390	W	Maximum loss across both freewheeling diodes
RECOMMENDED DIODE	UF1JLW		UF1JLW		Recommended freewheeling diode. Two of this diode in series must be implemented to pass 80% voltage derating and thermal requirements
BIAS/FEEDBACK PARAMETERS					
VF_BIAS			0.70	V	Forward voltage drop of the bias diode
RBIAS			2490	Ohms	Bias resistor
CBP			1.0	uF	BP pin capacitor
RFB			21000	Ohms	Feedback resistor
CFB			10	uF	Feedback capacitor
C_SOFTSTART			1-10	uF	If the output voltage is greater than 12 V or total output and system capacitance is greater than 100 uF, a soft start capacitor between 1uF and 10 uF is recommended
PLOSS_FEEDBACK			0.014	W	Maximum feedback component losses
OUTPUT CAPACITOR					
OUTPUT VOLTAGE RIPPLE			360	mV	Desired output voltage ripple
IRMS_COUT			0.163	A	Maximum output capacitor RMS current
PLOSS_COUT			0.031	W	Maximum output capacitor power loss
ESR_COUT			1153	mOhms	ESR of the output capacitor

Notes:

- 1) VDSOFF warning was addressed by using a Stack Fet Buck topology. Stack Fet Buck topology will limit the Vds of the LNK3206GQ to a safe level.
- 2) Freewheeling diode PIV info was addressed by using two 1000V rated diodes in series.



8 Performance Data

8.1 *Efficiency*

8.1.1 Line Efficiency

Test Condition: Soak for 10 minutes for each line.

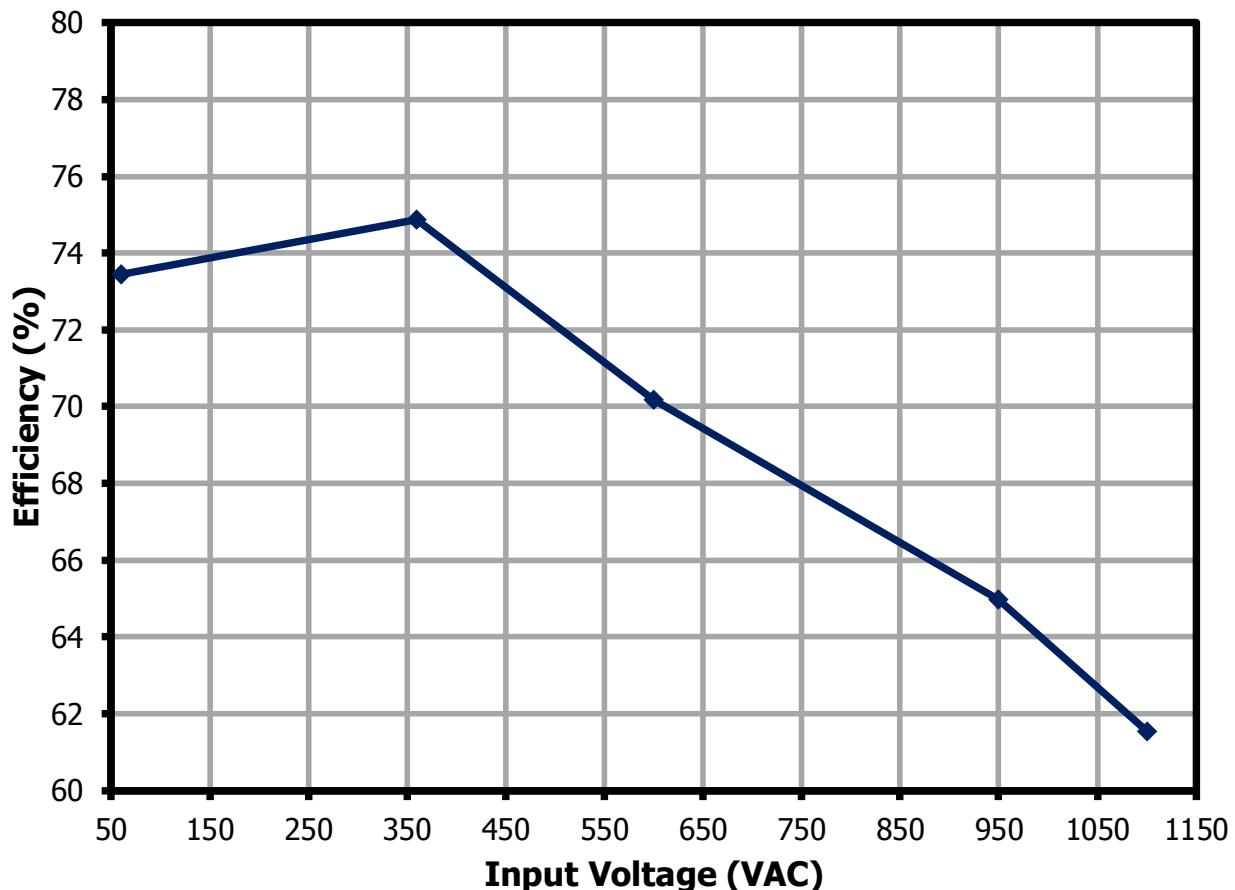


Figure 6 – Efficiency vs. Input Voltage.

8.1.2 Load Efficiency

Test Condition: Soak for 10 minutes, 5 minutes delay per each line, and 10 sec. delay for each load

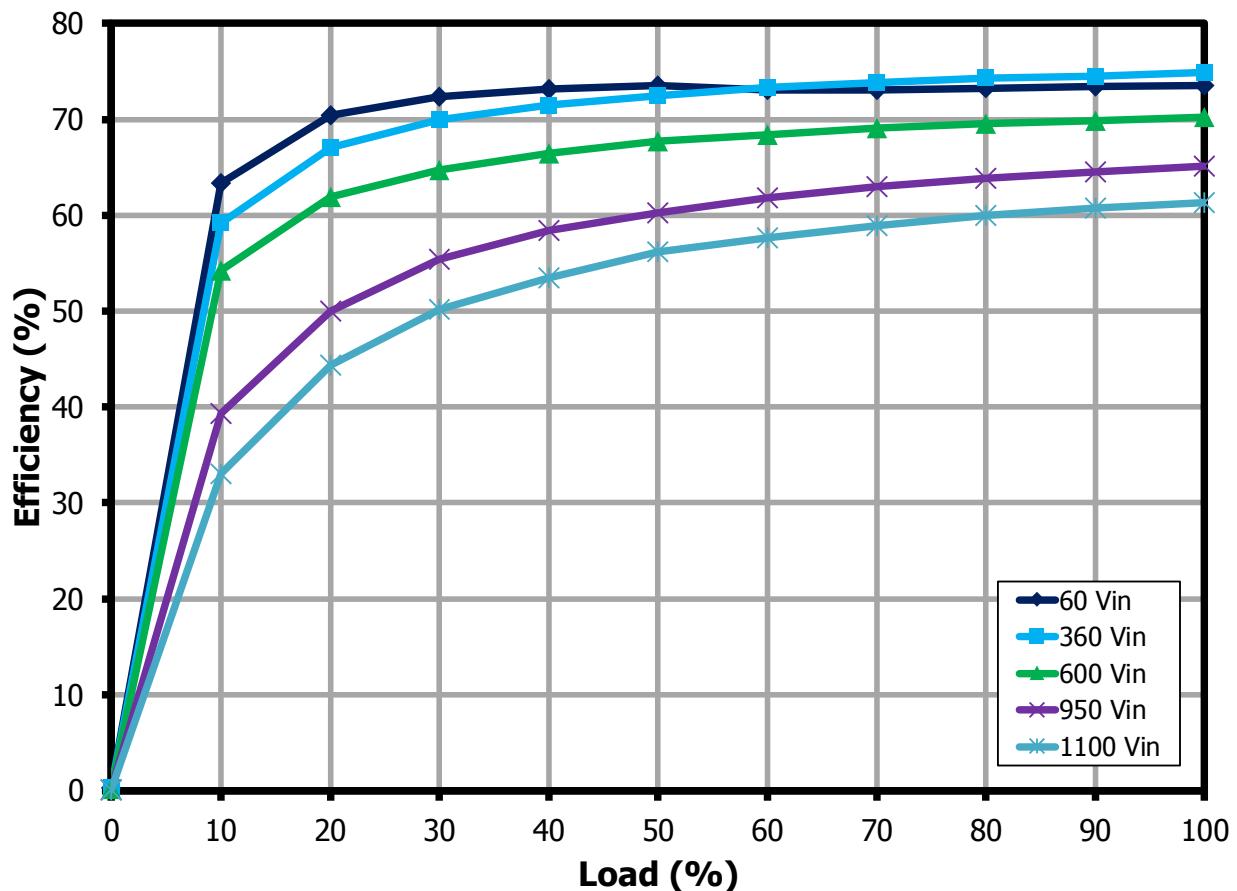


Figure 7 – Efficiency vs. Percentage Load.

8.2 ***Line Regulation***

Test Condition: Soak for 10 minutes for each line.

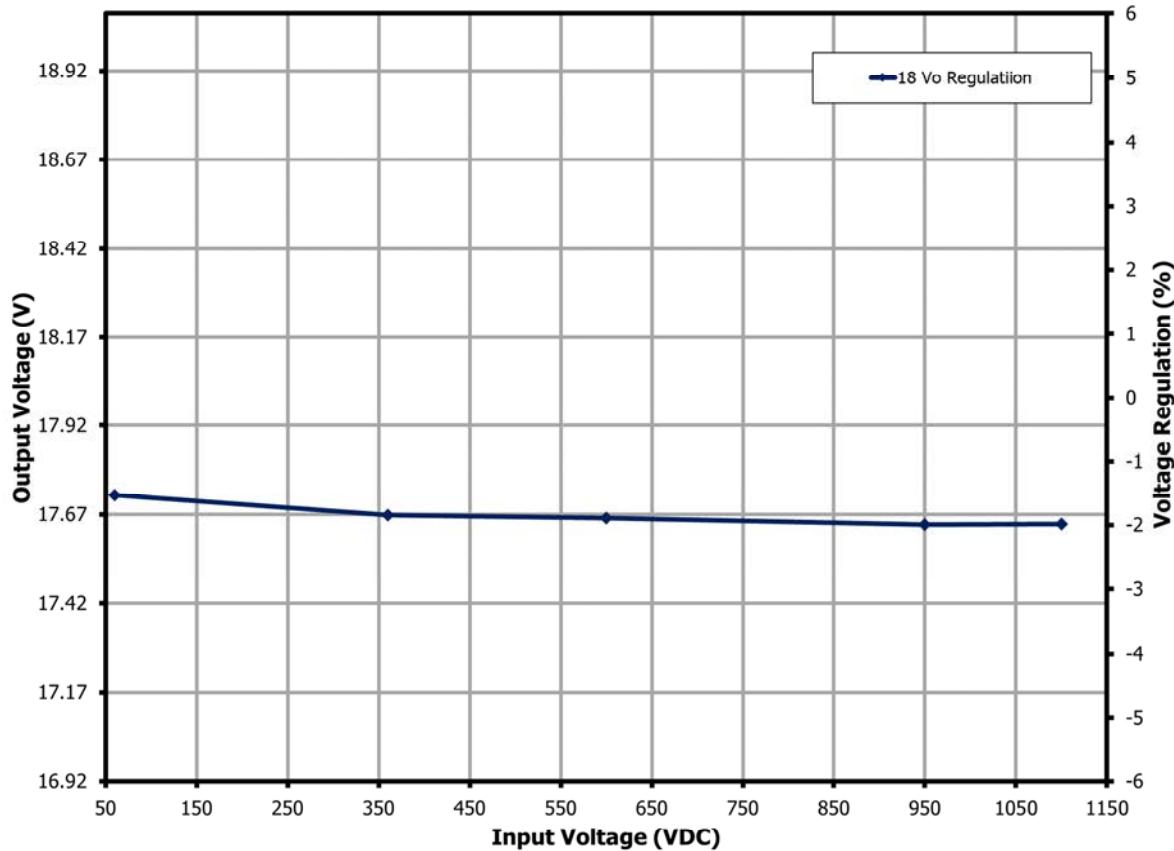


Figure 8 – Output Voltage vs. Input Voltage.



8.3 ***Load Regulation***

Test Condition: Soak for 10 minutes, 5 minutes delay per each line, and 10 sec. delay for each load

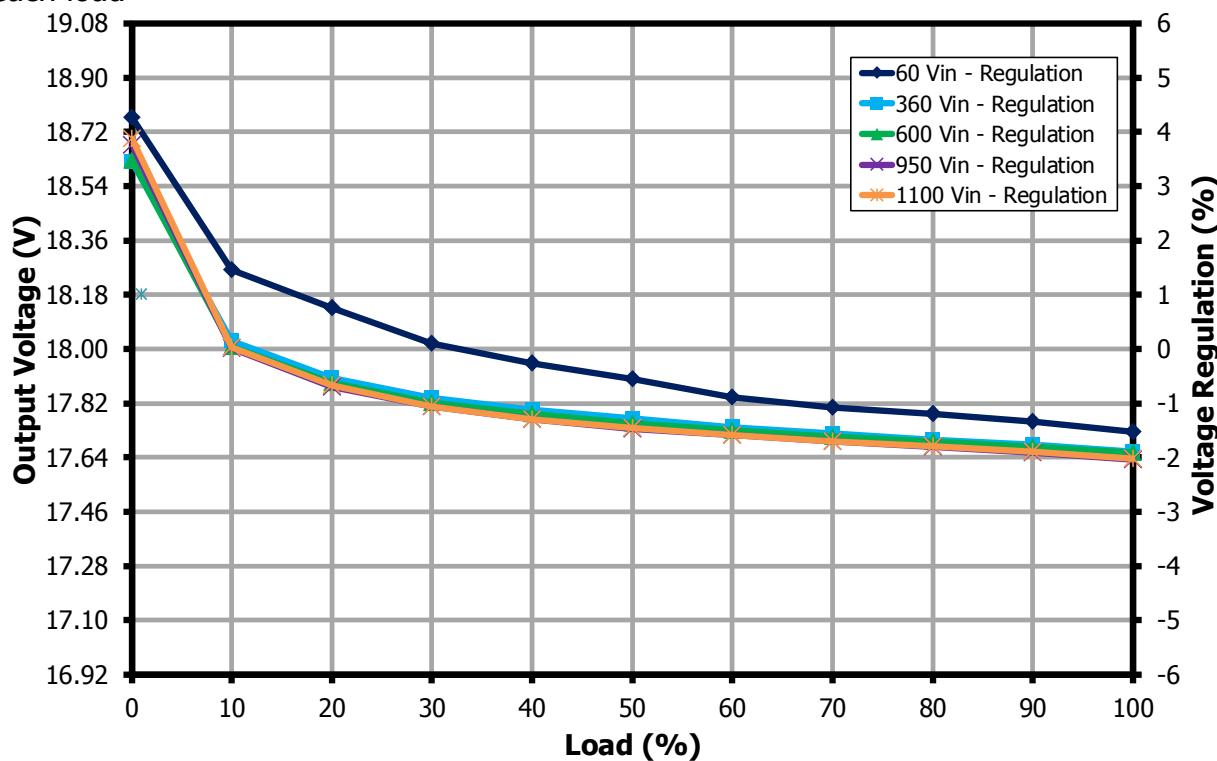


Figure 9 – Output Voltage vs. Percent Load.

9 Waveforms

9.1 Load Transient Response

Test Condition: Dynamic load frequency = 1 kHz, Duty cycle = 50 %
Slew Rate = 0.8 A / μ s

9.1.1 18 V Transient 0% - 100% Load Change

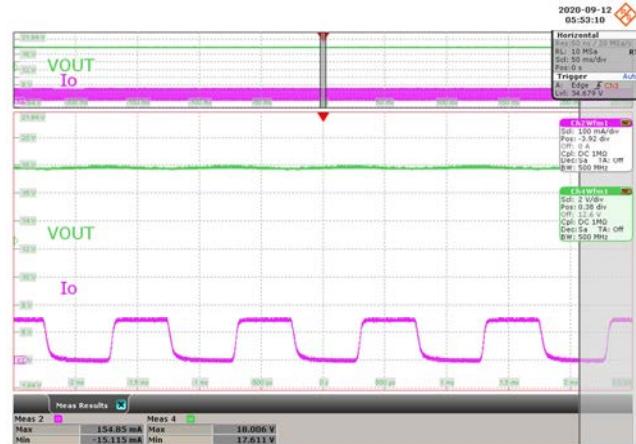
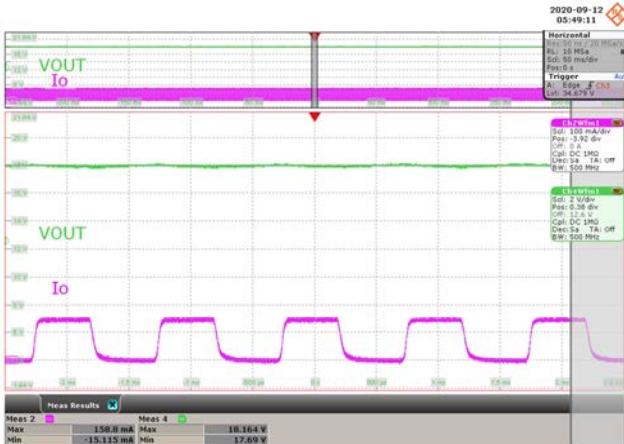


Figure 10 – 60 VDC Input.

CH2: I_{OUT} , 100 mA / div., 50 ms / div.
CH4: V_{OUT} , 2 V / div., 50 ms / div.
Zoom: 500 μ s / div.
 $V_{OUT(MAX)} = 18.164$ V, $V_{OUT(MIN)} = 17.69$ V.

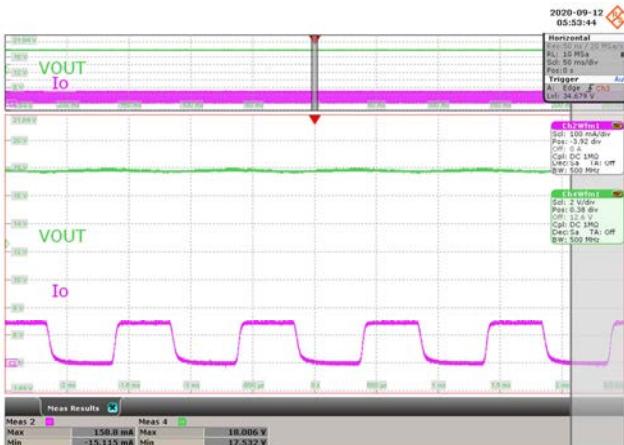


Figure 12 – 600 VDC Input.

CH2: I_{OUT} , 100 mA / div., 50 ms / div.
CH4: V_{OUT} , 2 V / div., 50 ms / div.
Zoom: 500 μ s / div.
 $V_{OUT(MAX)} = 18.006$ V, $V_{OUT(MIN)} = 17.532$ V.

Figure 11 – 360 VDC Input.

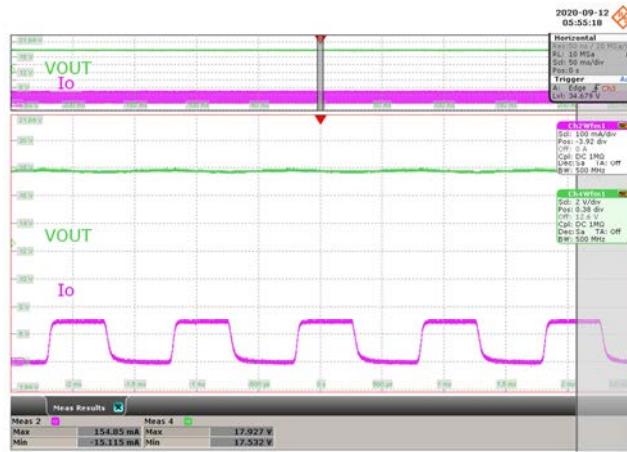
CH2: I_{OUT} , 100 mA / div., 50 ms / div.
CH4: V_{OUT} , 2 V / div., 50 ms / div.
Zoom: 500 μ s / div.
 $V_{OUT(MAX)} = 18.006$ V, $V_{OUT(MIN)} = 17.611$ V.



Figure 13 – 950 VDC Input.

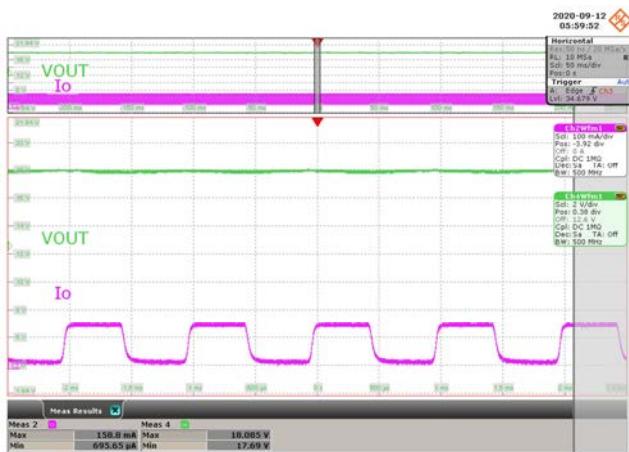
CH2: I_{OUT} , 100 mA / div., 50 ms / div.
CH4: V_{OUT} , 2 V / div., 50 ms / div.
Zoom: 500 μ s / div.
 $V_{OUT(MAX)} = 17.927$ V, $V_{OUT(MIN)} = 17.532$ V.



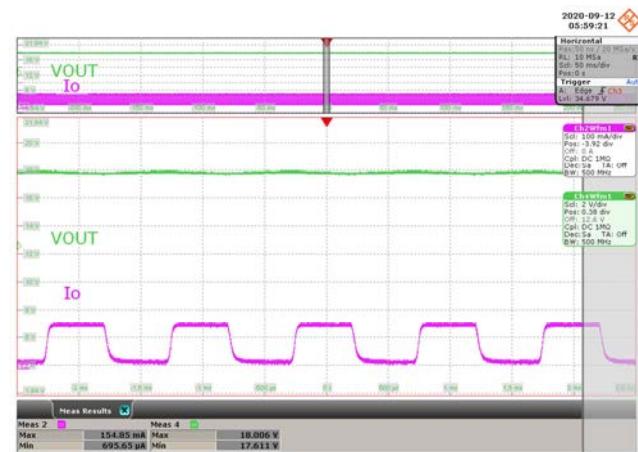
**Figure 14 – 1100 VDC Input.**CH2: I_{out} , 100 mA / div., 50 ms / div.CH4: V_{out} , 2 V / div., 50 ms / div.Zoom: 500 μ s / div.

$$V_{out(\text{MAX})} = 17.927 \text{ V}, V_{out(\text{MIN})} = 17.532 \text{ V.}$$

9.1.2 18 V Transient 10% - 100% Load Change

**Figure 15 – 60 VDC Input.**CH2: I_{out} , 100 mA / div., 50 ms / div.CH4: V_{out} , 2 V / div., 50 ms / div.Zoom: 500 μ s / div.

$$V_{out(\text{MAX})} = 18.085 \text{ V}, V_{out(\text{MIN})} = 17.69 \text{ V.}$$

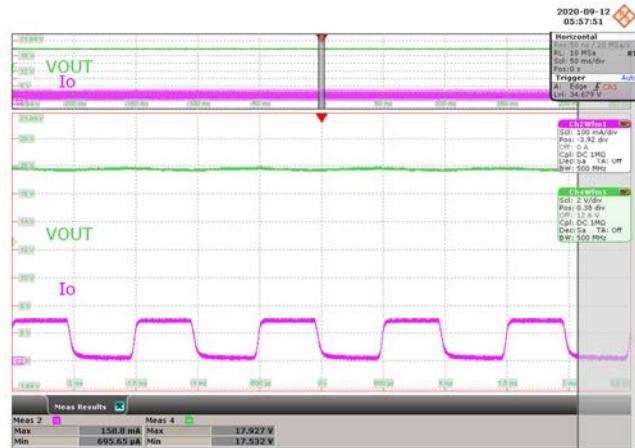
**Figure 16 – 360 VDC Input.**CH2: I_{out} , 100 mA / div., 50 ms / div.CH4: V_{out} , 2 V / div., 50 ms / div.Zoom: 500 μ s / div.

$$V_{out(\text{MAX})} = 18.006 \text{ V}, V_{out(\text{MIN})} = 17.611 \text{ V.}$$

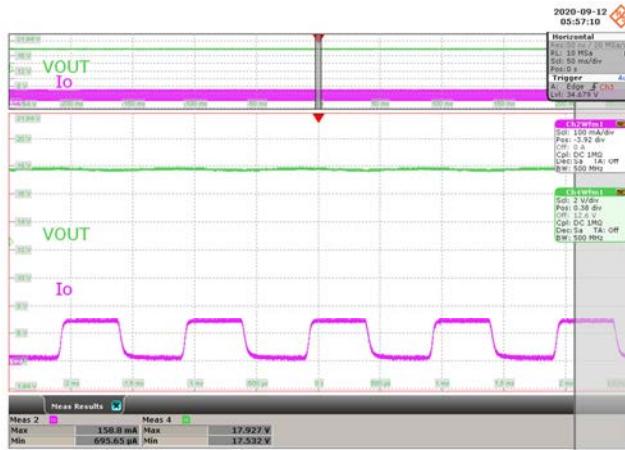


**Figure 17** – 600 VDC Input.

CH2: I_{OUT} , 100 mA / div., 50 ms / div.
 CH4: V_{OUT} , 2 V / div., 50 ms / div.
 Zoom: 500 μ s / div.
 $V_{OUT(MAX)} = 18.006$ V, $V_{OUT(MIN)} = 17.532$ V.

**Figure 18** – 950 VDC Input.

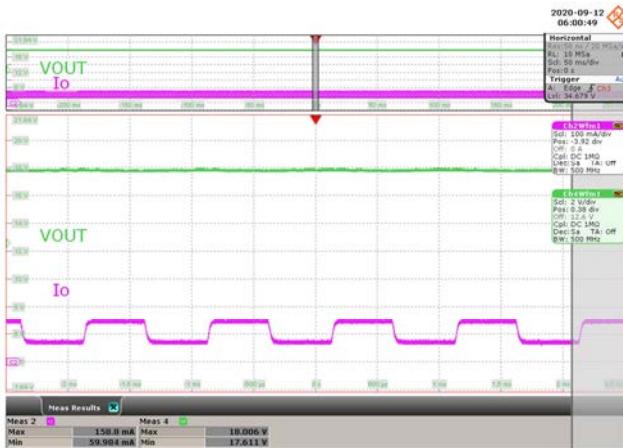
CH2: I_{OUT} , 100 mA / div., 50 ms / div.
 CH4: V_{OUT} , 2 V / div., 50 ms / div.
 Zoom: 500 μ s / div.
 $V_{OUT(MAX)} = 17.927$ V, $V_{OUT(MIN)} = 17.532$ V.

**Figure 19** – 1100 VDC Input.

CH2: I_{OUT} , 100 mA / div., 50 ms / div.
 CH4: V_{OUT} , 2 V / div., 50 ms / div.
 Zoom: 500 μ s / div.
 $V_{OUT(MAX)} = 17.927$ V, $V_{OUT(MIN)} = 17.532$ V.



9.1.3 18 V Transient 50% - 100% Load Change

**Figure 20** – 60 VDC Input.CH2: I_{OUT} , 100 mA / div., 50 ms / div.CH4: V_{OUT} , 2 V / div., 50 ms / div.Zoom: 500 μ s / div.

$$V_{\text{OUT}(\text{MAX})} = 18.006 \text{ V}, V_{\text{OUT}(\text{MIN})} = 17.611 \text{ V.}$$

**Figure 21** – 360 VDC Input.CH2: I_{OUT} , 100 mA / div., 50 ms / div.CH4: V_{OUT} , 2 V / div., 50 ms / div.Zoom: 500 μ s / div.

$$V_{\text{OUT}(\text{MAX})} = 17.927 \text{ V}, V_{\text{OUT}(\text{MIN})} = 17.532 \text{ V.}$$

**Figure 22** – 600 VDC Input.CH2: I_{OUT} , 100 mA / div., 50 ms / div.CH4: V_{OUT} , 2 V / div., 50 ms / div.Zoom: 500 μ s / div.

$$V_{\text{OUT}(\text{MAX})} = 17.927 \text{ V}, V_{\text{OUT}(\text{MIN})} = 17.532 \text{ V.}$$

**Figure 23** – 950 VDC Input.CH2: I_{OUT} , 100 mA / div., 50 ms / div.CH4: V_{OUT} , 2 V / div., 50 ms / div.Zoom: 500 μ s / div.

$$V_{\text{OUT}(\text{MAX})} = 17.848 \text{ V}, V_{\text{OUT}(\text{MIN})} = 17.532 \text{ V.}$$





Figure 24 – 1100 VDC Input.

CH2: I_{OUT} , 100 mA / div., 50 ms / div.

CH4: V_{OUT} , 2 V / div., 50 ms / div.

Zoom: 500 μ s / div.

$V_{\text{OUT}(\text{MAX})} = 17.848 \text{ V}$, $V_{\text{OUT}(\text{MIN})} = 17.532 \text{ V}$.



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9.2 Switching Waveforms

9.2.1 Primary MOSFET V_{DS} and I_{DS} at Normal Operation

9.2.1.1 Full Load

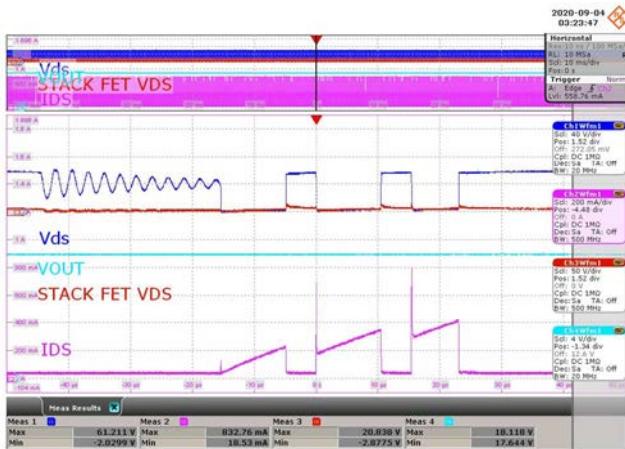


Figure 25 – 60 VDC Input.

CH1: IC V_{DS} , 40 V / div., 10 ms / div.
 CH2: I_{DS} , 200 mA / div., 10 ms / div.
 CH3: FET V_{DS} , 50 V / div., 10 ms / div.
 CH4: V_{OUT} , 4 V / div., 10 ms / div.

IC $V_{DS(MAX)}$ = 61.21 V, $I_{DS(MAX)}$ = 832.76 mA.
 Stack FET $V_{DS(MAX)}$ = 20.83 V.

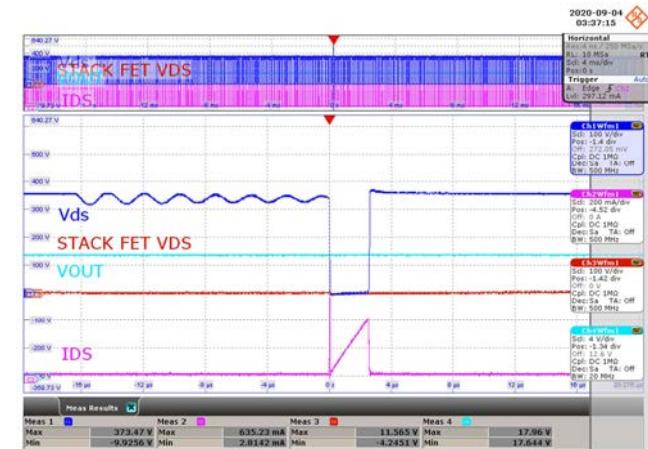


Figure 26 – 360 VDC Input.

CH1: IC V_{DS} , 100 V / div., 4 ms / div.
 CH2: I_{DS} , 200 mA / div., 4 ms / div.
 CH3: FET V_{DS} , 100 V / div., 4 ms / div.
 CH4: V_{OUT} , 4 V / div., 4 ms / div.

IC $V_{DS(MAX)}$ = 373.47 V, $I_{DS(MAX)}$ = 635.23 mA.
 Stack FET $V_{DS(MAX)}$ = 11.56 V.

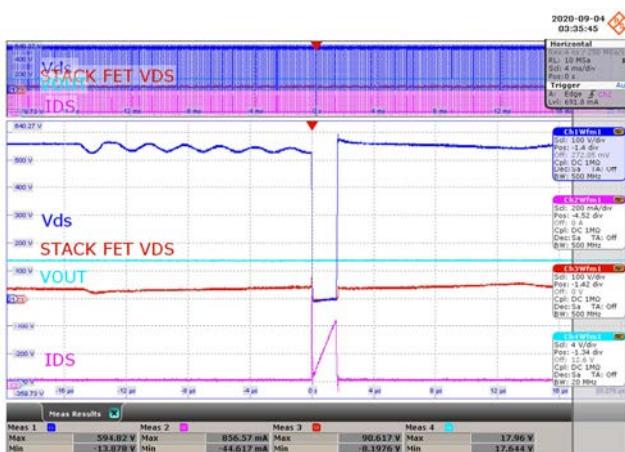


Figure 27 – 600 VDC Input.

CH1: IC V_{DS} , 100 V / div., 4 ms / div.
 CH2: I_{DS} , 200 mA / div., 4 ms / div.
 CH3: FET V_{DS} , 100 V / div., 4 ms / div.
 CH4: V_{OUT} , 4 V / div., 4 ms / div.

IC $V_{DS(MAX)}$ = 594.82 V, $I_{DS(MAX)}$ = 856.57 mA.
 Stack FET $V_{DS(MAX)}$ = 90.617 V.

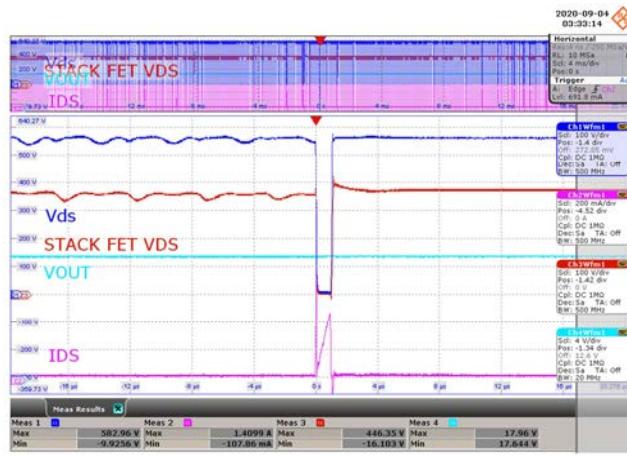


Figure 28 – 950 VDC Input.

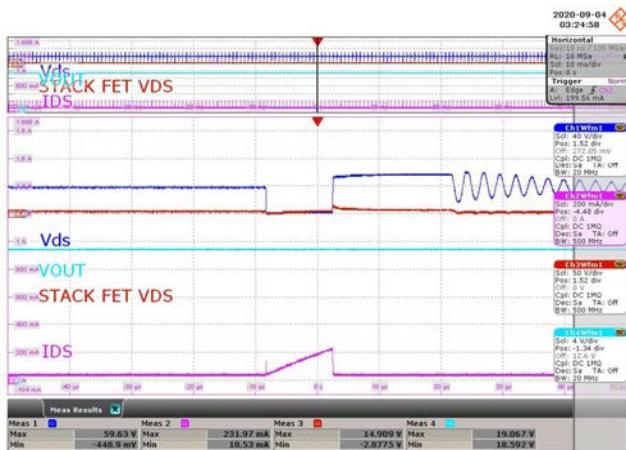
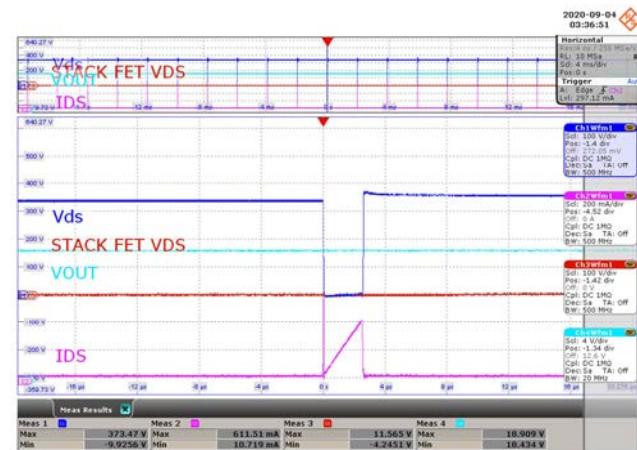
CH1: IC V_{DS} , 100 V / div., 4 ms / div.
 CH2: I_{DS} , 200 mA / div., 4 ms / div.
 CH3: FET V_{DS} , 100 V / div., 4 ms / div.
 CH4: V_{OUT} , 4 V / div., 4 ms / div.

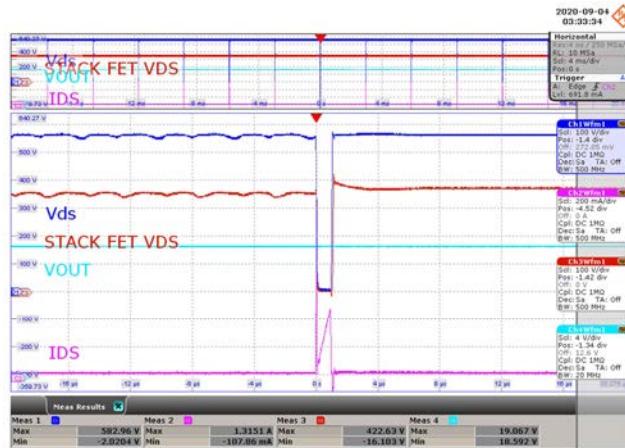
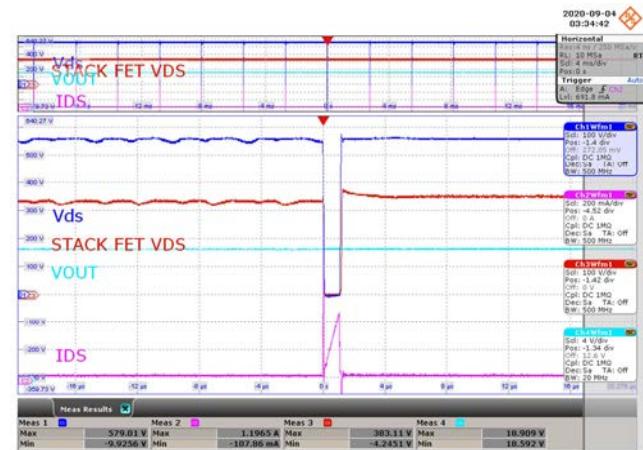
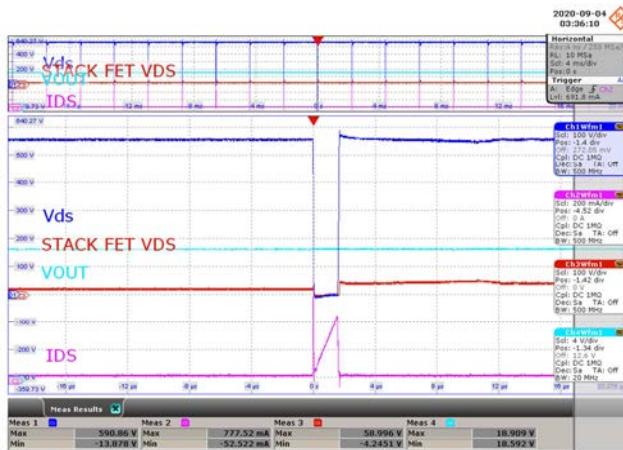
IC $V_{DS(MAX)}$ = 582.96 V, $I_{DS(MAX)}$ = 1.23 A.
 Stack FET $V_{DS(MAX)}$ = 434.49 V.



**Figure 29** – 1100 VDC Input.CH1: IC V_{DS} , 100 V / div., 4 ms / div.CH2: I_{DS} , 200 mA / div., 4 ms / div.CH3: FET V_{DS} , 100 V / div., 4 ms / div.CH4: V_{OUT} , 4 V / div., 4 ms / div.IC $V_{DS(\text{MAX})} = 582.96$ V, $I_{DS(\text{MAX})} = 1.40$ A.Stack FET $V_{DS(\text{MAX})} = 446.35$ V.

9.2.1.2 No-Load

**Figure 30** – 60 VDC Input.CH1: IC V_{DS} , 40 V / div., 10 ms / div.CH2: I_{DS} , 200 mA / div., 10 ms / div.CH3: FET V_{DS} , 50 V / div., 10 ms / div.CH4: V_{OUT} , 4 V / div., 10 ms / div.IC $V_{DS(\text{MAX})} = 59.63$ V, $I_{DS(\text{MAX})} = 231.97$ mA.Stack FET $V_{DS(\text{MAX})} = 14.909$ V.**Figure 31** – 360 VDC Input.CH1: IC V_{DS} , 100 V / div., 4 ms / div.CH2: I_{DS} , 200 mA / div., 4 ms / div.CH3: FET V_{DS} , 100 V / div., 4 ms / div.CH4: V_{OUT} , 4 V / div., 4 ms / div.IC $V_{DS(\text{MAX})} = 373.47$ V, $I_{DS(\text{MAX})} = 611.51$ mA.Stack FET $V_{DS(\text{MAX})} = 11.565$ V.



9.2.2 Primary MOSFET V_{DS} and I_{DS} at Start-up Operation

9.2.2.1 Full Load

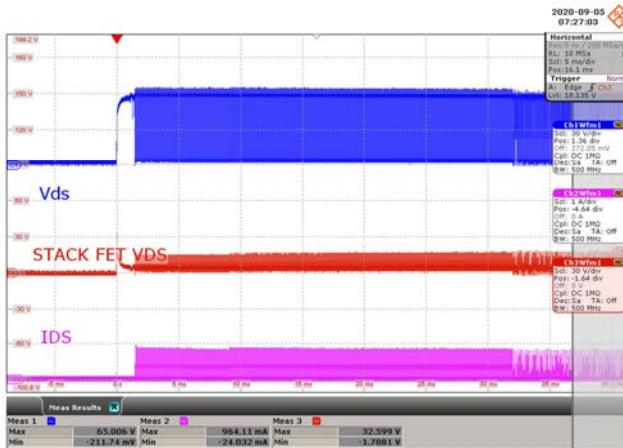


Figure 35 – 60 VDC Input.

CH1: IC V_{DS} , 30 V / div., 5 ms / div.

Ch2: I_{DS} , 1 A / div., 5 ms / div.

CH3: FET V_{DS} , 30 V / div., 5 ms / div.

IC $V_{DS(MAX)} = 65.00$ V, $I_{DS(MAX)} = 964.11$ mA.

Stack FET $V_{DS(MAX)} = 32.59$ V.

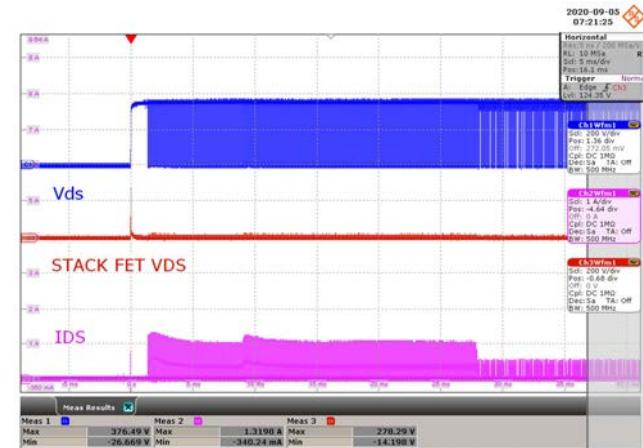


Figure 36 – 360 VDC Input.

CH1: IC V_{DS} , 200 V / div., 5 ms / div.

Ch2: I_{DS} , 1 A / div., 5 ms / div.

CH3: FET V_{DS} , 200 V / div., 5 ms / div.

IC $V_{DS(MAX)} = 376.49$ V, $I_{DS(MAX)} = 1.3198$ A.

Stack FET $V_{DS(MAX)} = 278.29$ V.

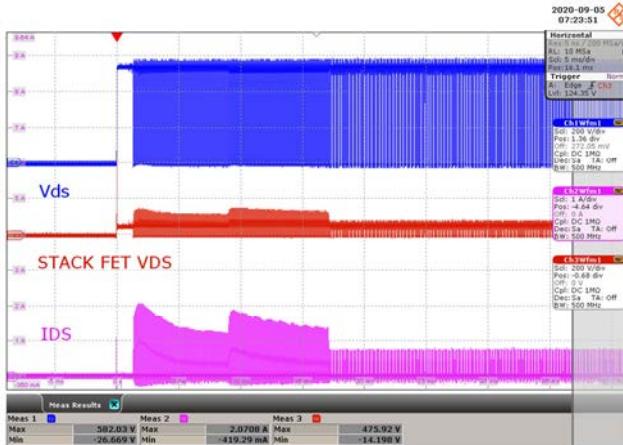


Figure 37 – 600 VDC Input.

CH1: IC V_{DS} , 200 V / div., 5 ms / div.

Ch2: I_{DS} , 1 A / div., 5 ms / div.

CH3: FET V_{DS} , 200 V / div., 5 ms / div.

IC $V_{DS(MAX)} = 582.03$ V, $I_{DS(MAX)} = 2.07$ A.

Stack FET $V_{DS(MAX)} = 475.92$ V.

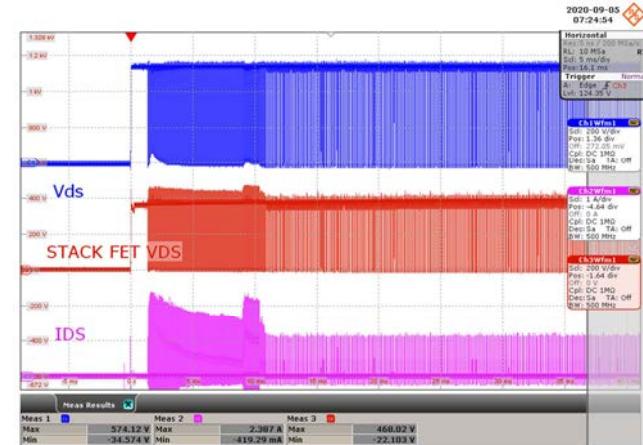


Figure 38 – 950 VDC Input.

CH1: IC V_{DS} , 200 V / div., 5 ms / div.

Ch2: I_{DS} , 1 A / div., 5 ms / div.

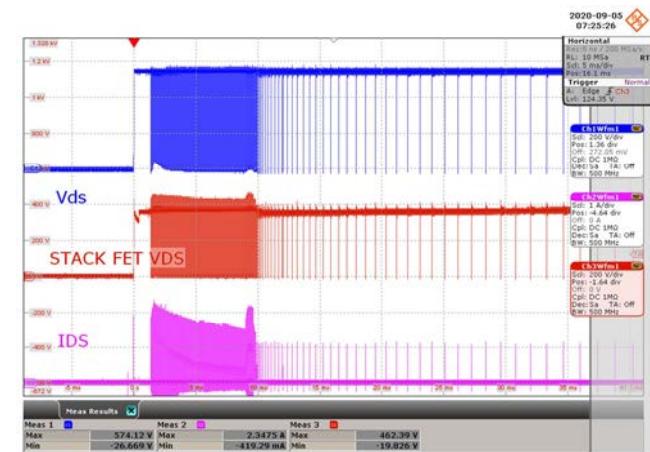
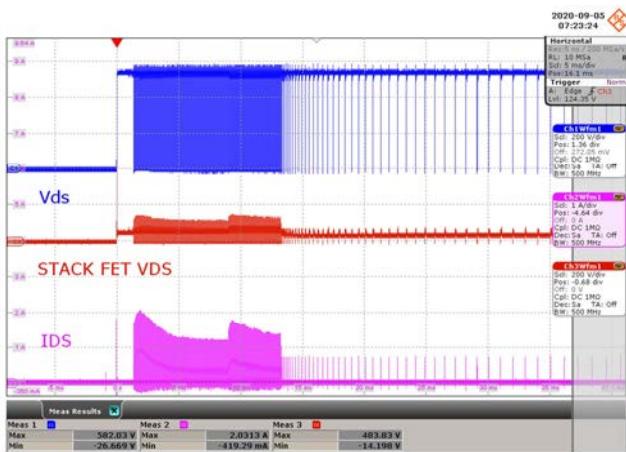
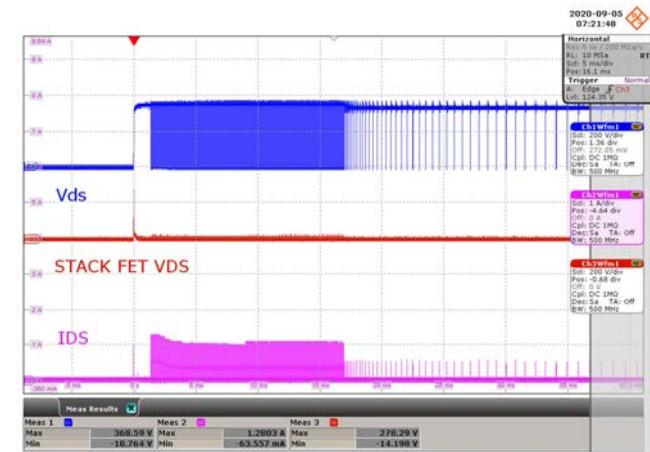
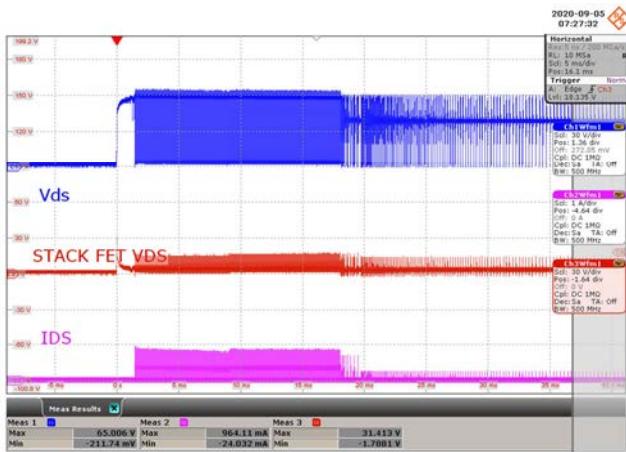
CH3: FET V_{DS} , 200 V / div., 5 ms / div.

IC $V_{DS(MAX)} = 574.12$ V, $I_{DS(MAX)} = 2.38$ A.

Stack FET $V_{DS(MAX)} = 468.02$ V.

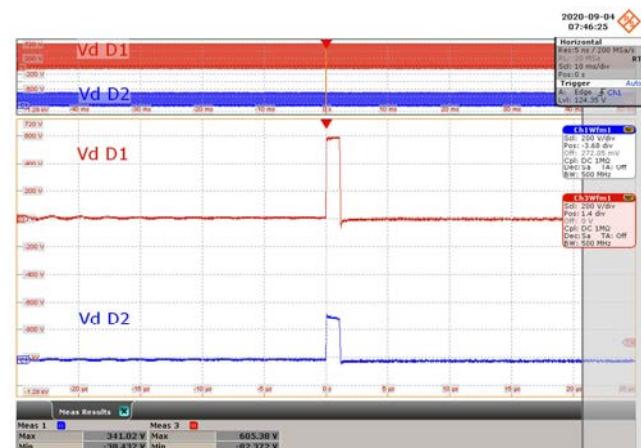
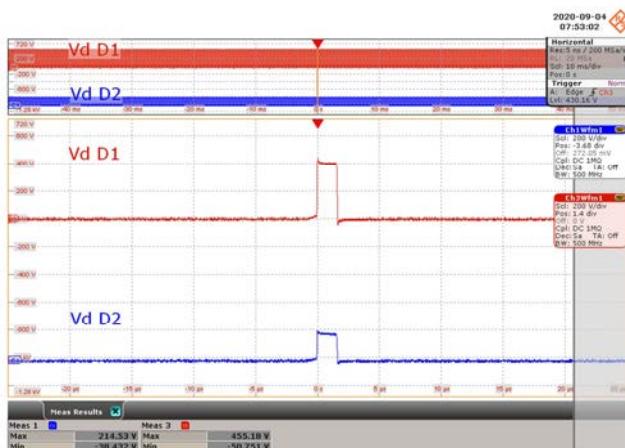
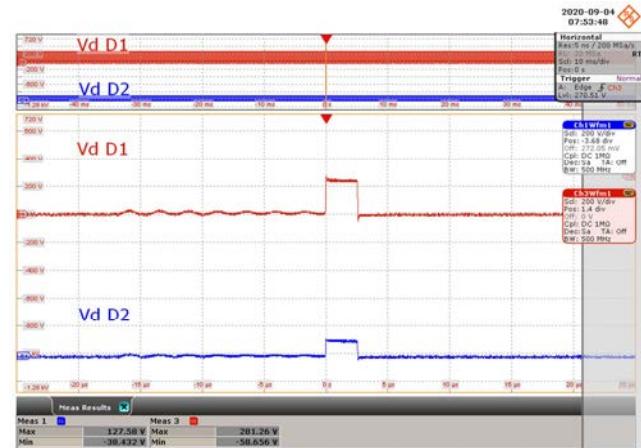
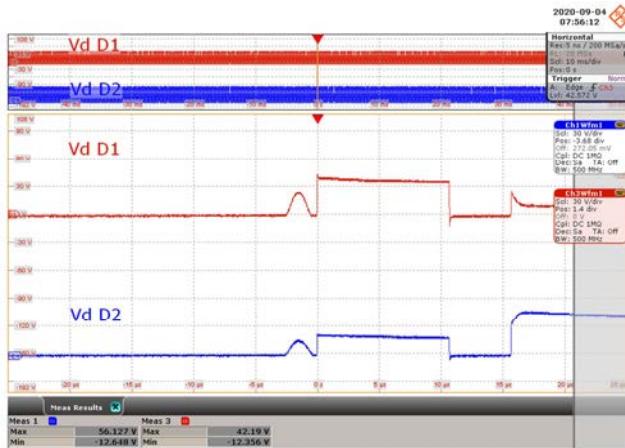


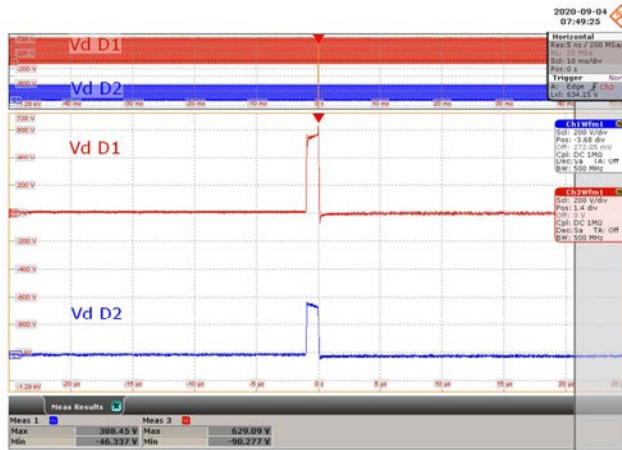
9.2.2.2 No-Load



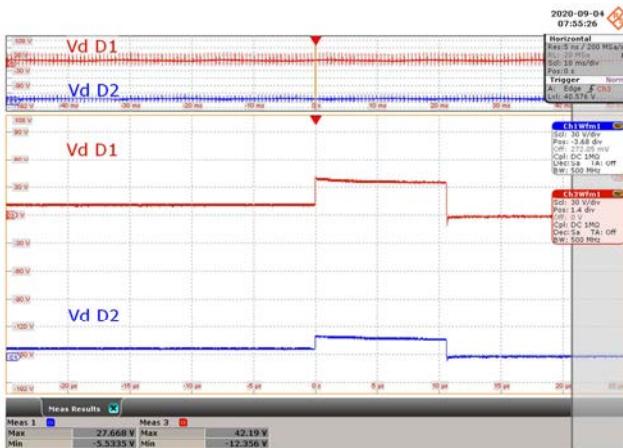
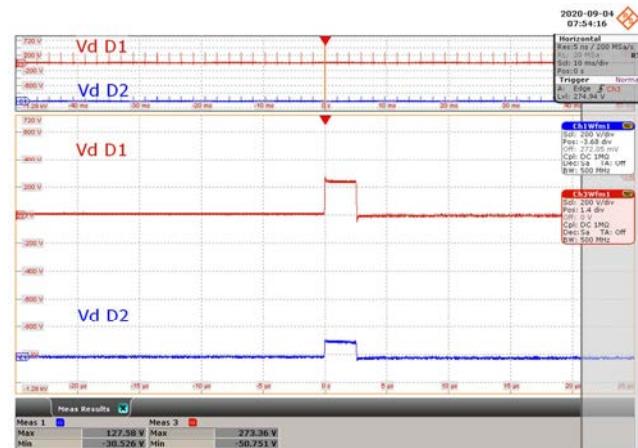
9.2.3 Free Wheeling Diode Voltage at Normal Operation

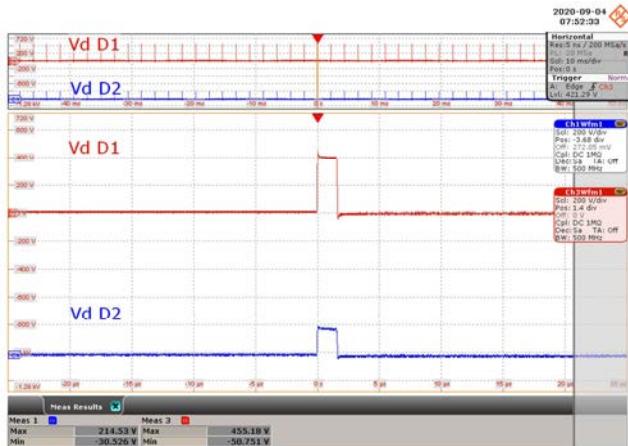
9.2.3.1 Full Load



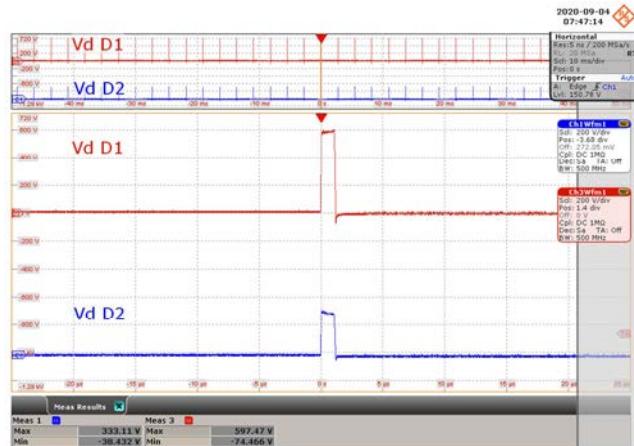
**Figure 47** – 1100 VDC Input.CH1: $V_{R,DIODE2}$, 200 V / div., 10 ms / div.CH3: $V_{R,DIODE1}$, 200 V / div., 10 ms / div.Zoom: 5 μ s / div. $PIV_{D1} = 629.09$ V, $PIV_{D2} = 388.45$ V.

9.2.3.2 No-Load

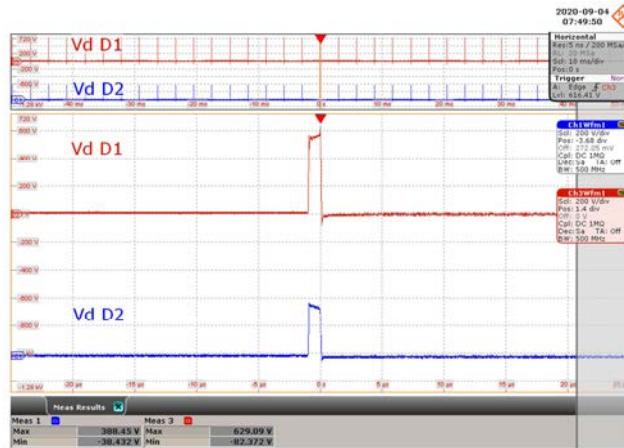
**Figure 48** – 60 VDC Input.CH1: $V_{R,DIODE2}$, 30 V / div., 10 ms / div.CH3: $V_{R,DIODE1}$, 30 V / div., 10 ms / div.Zoom: 5 μ s / div. $PIV_{D1} = 42.19$ V, $PIV_{D2} = 27.66$ V.**Figure 49** – 360 VDC Input.CH1: $V_{R,DIODE2}$, 200 V / div., 10 ms / div.CH3: $V_{R,DIODE1}$, 200 V / div., 10 ms / div.Zoom: 5 μ s / div. $PIV_{D1} = 273.36$ V, $PIV_{D2} = 127.58$ V.

**Figure 50** – 600 VDC Input.CH1: $V_{R,DIODE2}$, 200 V / div., 10 ms / div.CH3: $V_{R,DIODE1}$, 200 V / div., 10 ms / div.Zoom: 5 μ s / div.

$$PIV_{D1} = 455.18 \text{ V}, PIV_{D2} = 214.53 \text{ V.}$$

**Figure 51** – 950 VDC Input.CH1: $V_{R,DIODE2}$, 200 V / div., 10 ms / div.CH3: $V_{R,DIODE1}$, 200 V / div., 10 ms / div.Zoom: 5 μ s / div.

$$PIV_{D1} = 597.47 \text{ V}, PIV_{D2} = 333.11 \text{ V.}$$

**Figure 52** – 1100 VDC Input.CH1: $V_{R,DIODE2}$, 200 V / div., 10 ms / div.CH3: $V_{R,DIODE1}$, 200 V / div., 10 ms / div.Zoom: 5 μ s / div.

$$PIV_{D1} = 629.09 \text{ V}, PIV_{D2} = 388.45 \text{ V.}$$



9.3 ***Output Start-Up***

9.3.1 Output Start-Up Full Load CC

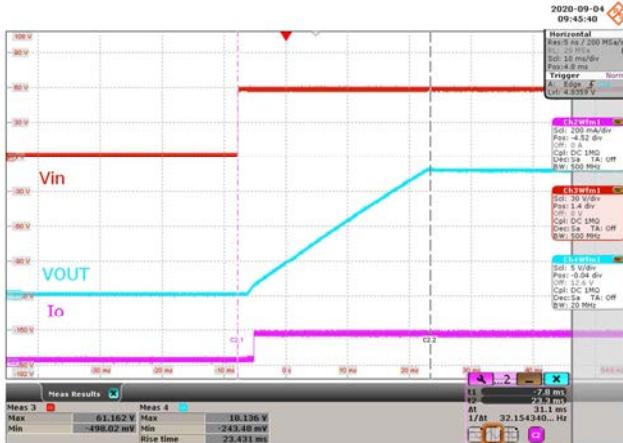


Figure 53 – 60 VDC Input.

Ch2: I_{OUT} , 200 mA / div., 10 ms / div.

CH3: V_{IN} , 30 V / div., 10 ms / div.

Ch4: V_{OUT} , 5 V / div., 10 ms / div.

$V_{OUT(MAX)} = 18.13 \text{ V}$, $V_{OUT(RISE)} = 23.43 \text{ ms}$.

$T_{V_O_to\ Vin} = 31.1 \text{ ms}$.

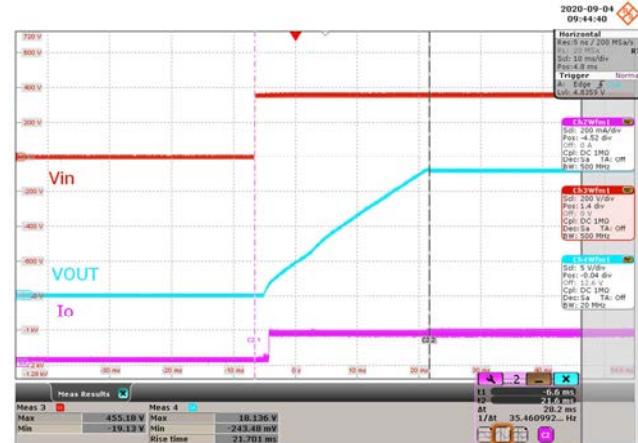


Figure 54 – 360 VDC Input.

Ch2: I_{OUT} , 200 mA / div., 10 ms / div.

CH3: V_{IN} , 200 V / div., 10 ms / div.

Ch4: V_{OUT} , 5 V / div., 10 ms / div.

$V_{OUT(MAX)} = 18.13 \text{ V}$, $V_{OUT(RISE)} = 21.70 \text{ ms}$.

$T_{V_O_to\ Vin} = 28.2 \text{ ms}$.

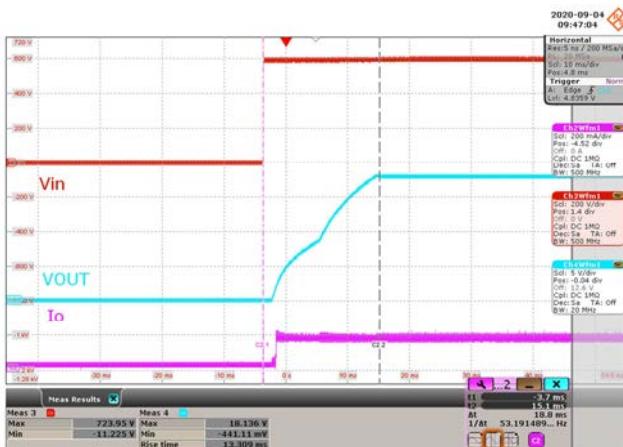


Figure 55 – 600 VDC Input.

Ch2: I_{OUT} , 200 mA / div., 10 ms / div.

CH3: V_{IN} , 200 V / div., 10 ms / div.

Ch4: V_{OUT} , 5 V / div., 10 ms / div.

$V_{OUT(MAX)} = 18.13 \text{ V}$, $V_{OUT(RISE)} = 13.30 \text{ ms}$.

$T_{V_O_to\ Vin} = 18.8 \text{ ms}$.

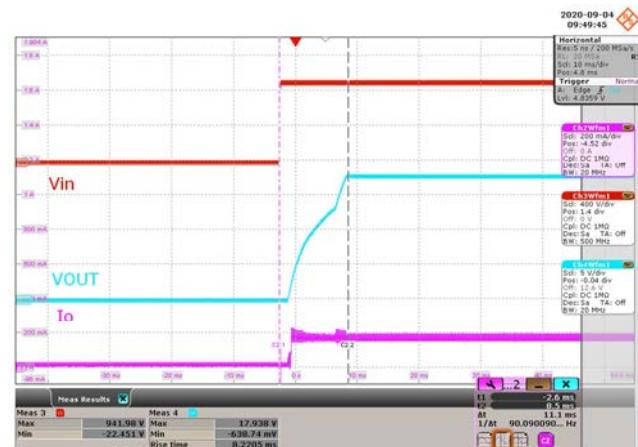


Figure 56 – 950 VDC Input.

Ch2: I_{OUT} , 200 mA / div., 10 ms / div.

CH3: V_{IN} , 400 V / div., 10 ms / div.

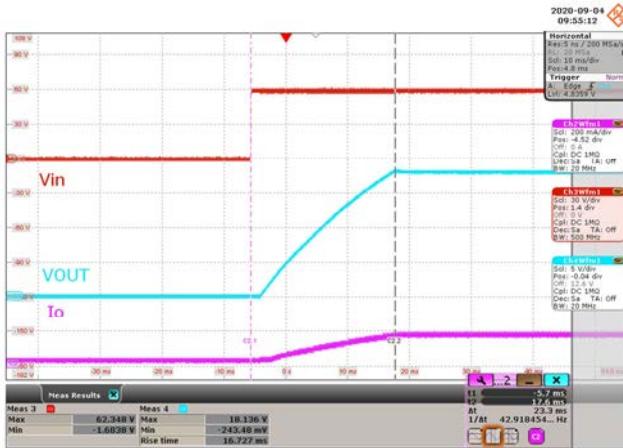
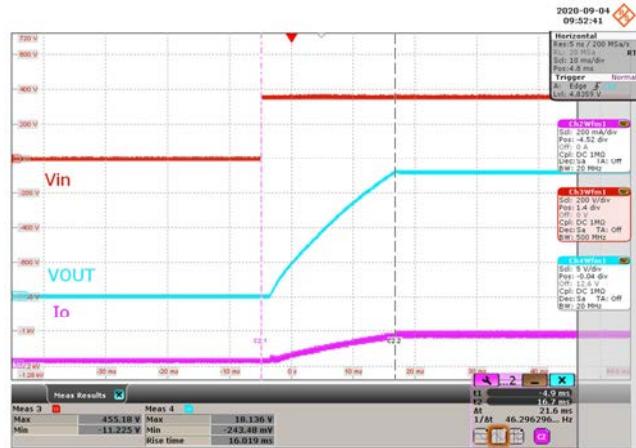
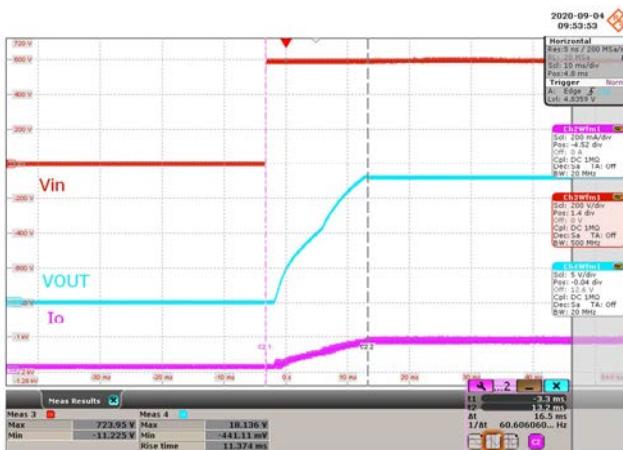
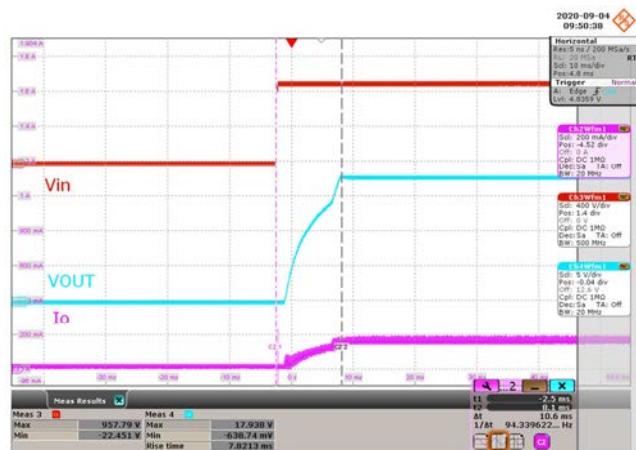
Ch4: V_{OUT} , 5 V / div., 10 ms / div.

$V_{OUT(MAX)} = 17.93 \text{ V}$, $V_{OUT(RISE)} = 8.22 \text{ ms}$.

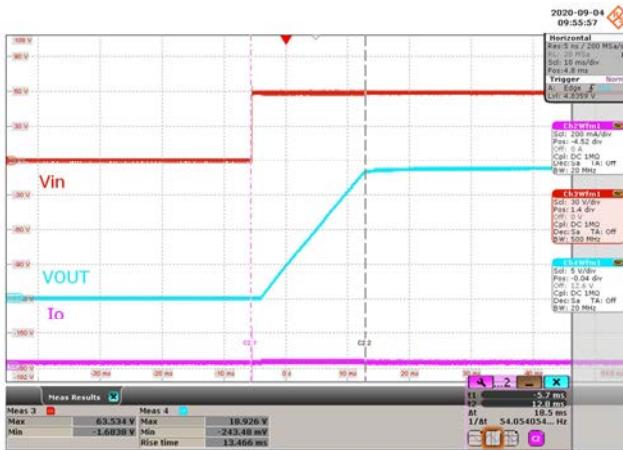
$T_{V_O_to\ Vin} = 11.1 \text{ ms}$.



9.3.2 Output Start-Up Full Load CR

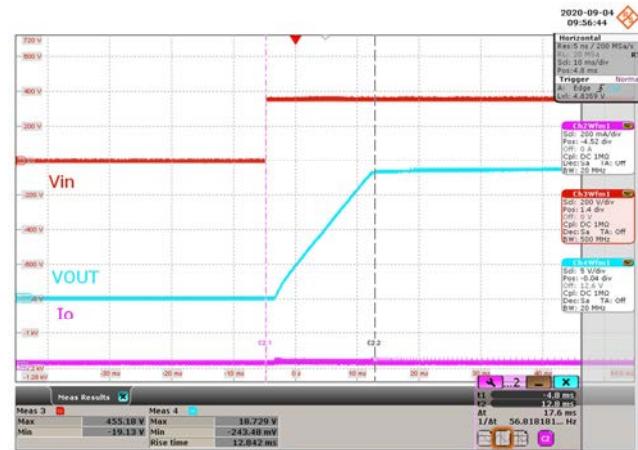
**Figure 57** – 60 VDC Input.**Figure 58** – 360 VDC Input.**Figure 59** – 600 VDC Input.**Figure 60** – 950 VDC Input.

9.3.3 Output Start-Up No-Load

**Figure 61** – 60 VDC Input.Ch2: I_{OUT} , 200 mA / div., 10 ms / div.CH3: V_{IN} , 30 V / div., 10 ms / div.Ch4: V_{OUT} , 5 V / div., 10 ms / div.

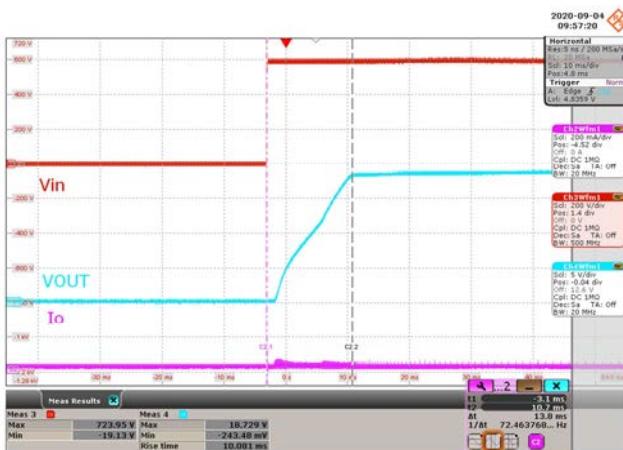
$V_{OUT(MAX)} = 18.92 \text{ V}, V_{OUT(RISE)} = 13.46 \text{ ms}.$

$T_{V_o_to_V_{in}} = 18.5 \text{ ms}.$

**Figure 62** – 360 VDC Input.Ch2: I_{OUT} , 200 mA / div., 10 ms / div.CH3: V_{IN} , 200 V / div., 10 ms / div.Ch4: V_{OUT} , 5 V / div., 10 ms / div.

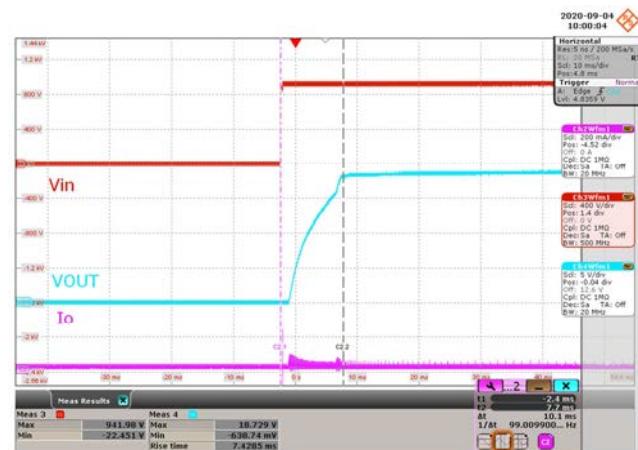
$V_{OUT(MAX)} = 18.72 \text{ V}, V_{OUT(RISE)} = 12.842 \text{ ms}.$

$T_{V_o_to_V_{in}} = 17.6 \text{ ms}.$

**Figure 63** – 600 VDC Input.Ch2: I_{OUT} , 200 mA / div., 10 ms / div.CH3: V_{IN} , 200 V / div., 10 ms / div.Ch4: V_{OUT} , 5 V / div., 10 ms / div.

$V_{OUT(MAX)} = 18.72 \text{ V}, V_{OUT(RISE)} = 10.08 \text{ ms}.$

$T_{V_o_to_V_{in}} = 13.8 \text{ ms}.$

**Figure 64** – 950 VDC Input.Ch2: I_{OUT} , 200 mA / div., 10 ms / div.CH3: V_{IN} , 400 V / div., 10 ms / div.Ch4: V_{OUT} , 5 V / div., 10 ms / div.

$V_{OUT(MAX)} = 18.72 \text{ V}, V_{OUT(RISE)} = 7.42 \text{ ms}.$

$T_{V_o_to_V_{in}} = 10.1 \text{ ms}.$



9.4 Fault Conditions

9.4.1 18 V Output Short-Circuit



Figure 65 – 60 VDC Input. Output Short.

CH1: IC V_{DS} , 30 V / div., 1 s / div.

Ch2: I_{DS} , 1 A / div., 1 s / div.

CH3: FET V_{DS} , 30 V / div., 1 s / div.

IC $V_{DS(\text{MAX})} = 66.19$ V., $I_{DS(\text{MAX})} = 924.58$ mA.

Stack FET $V_{DS(\text{MAX})} = 30.22$ V.

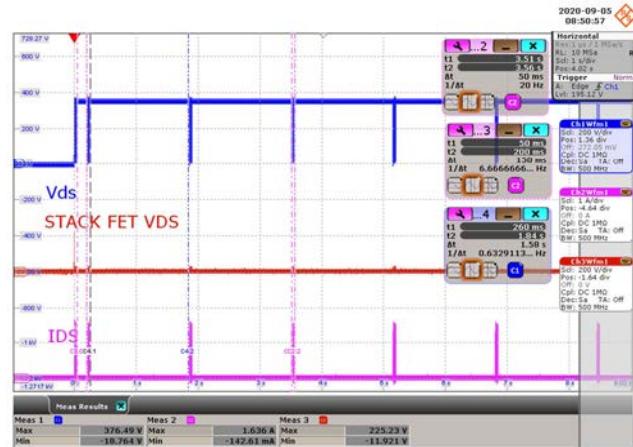


Figure 66 – 360 VDC Input. Output Short.

CH1: IC V_{DS} , 200 V / div., 1 s / div.

Ch2: I_{DS} , 1 A / div., 1 s / div.

CH3: FET V_{DS} , 200 V / div., 1 s / div.

IC $V_{DS(\text{MAX})} = 376.49$ V., $I_{DS(\text{MAX})} = 1.63$ A.

Stack FET $V_{DS(\text{MAX})} = 225.53$ V.



Figure 67 – 600 VDC Input. Output Short.

CH1: IC V_{DS} , 200 V / div., 1 s / div.

Ch2: I_{DS} , 1 A / div., 1 s / div.

CH3: FET V_{DS} , 200 V / div., 1 s / div.

IC $V_{DS(\text{MAX})} = 582.03$ V., $I_{DS(\text{MAX})} = 2.22$ A.

Stack FET $V_{DS(\text{MAX})} = 414.96$ V.

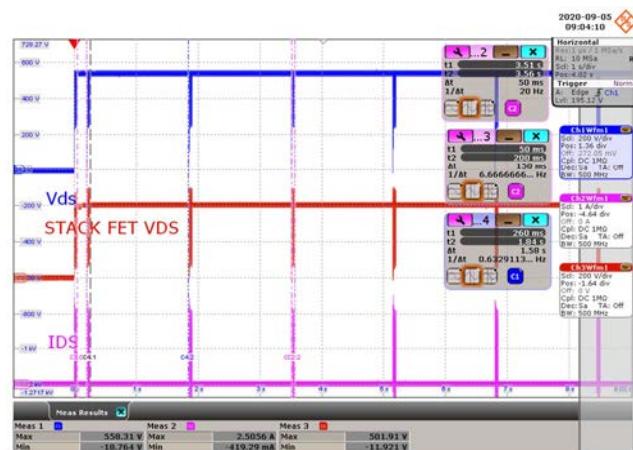


Figure 68 – 950 VDC Input. Output Short.

CH1: IC V_{DS} , 200 V / div., 1 s / div.

Ch2: I_{DS} , 1 A / div., 1 s / div.

CH3: FET V_{DS} , 200 V / div., 1 s / div.

IC $V_{DS(\text{MAX})} = 558.31$ V., $I_{DS(\text{MAX})} = 2.50$ A.

Stack FET $V_{DS(\text{MAX})} = 501.91$ V.



**Figure 69 – 1100 VDC Input. Output Short.**CH1: IC V_{DS} , 200 V / div., 1 s / div.Ch2: I_{DS} , 1 A / div., 1 s / div.CH3: FET V_{DS} , 200 V / div., 1 s / div.IC $V_{DS(\text{MAX})} = 574.12 \text{ V.}$, $I_{DS(\text{MAX})} = 2.14 \text{ A.}$ Stack FET $V_{DS(\text{MAX})} = 628.4 \text{ V.}$ 

9.5 ***Output Voltage Ripple***

9.5.1 Ripple Measurement Technique

For DC output ripple measurements, a modified oscilloscope test probe must be utilized in order to reduce spurious signals due to pick-up. Details of the probe modification are provided in the Figures below.

The 4987BA probe adapter is affixed with two capacitors tied in parallel across the probe tip. The capacitors include one (1) 0.1 μF / 50 V ceramic type and one (1) 47 μF / 50 V aluminum electrolytic. The aluminum electrolytic type capacitor is polarized, so proper polarity across DC outputs must be maintained (see below).

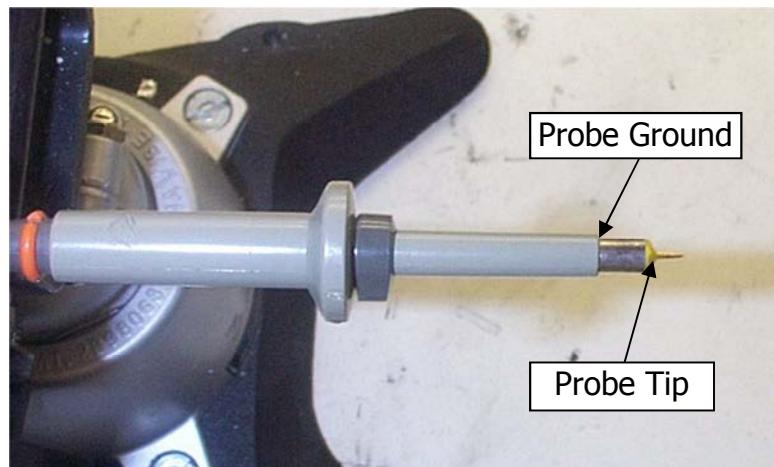


Figure 70 – Oscilloscope Probe Prepared for Ripple Measurement. (End Cap and Ground Lead Removed.)

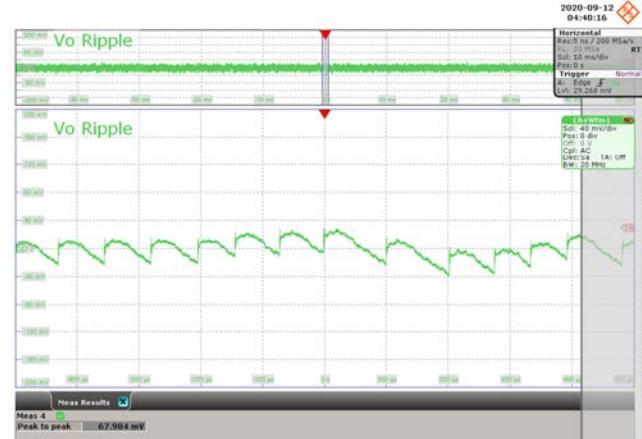
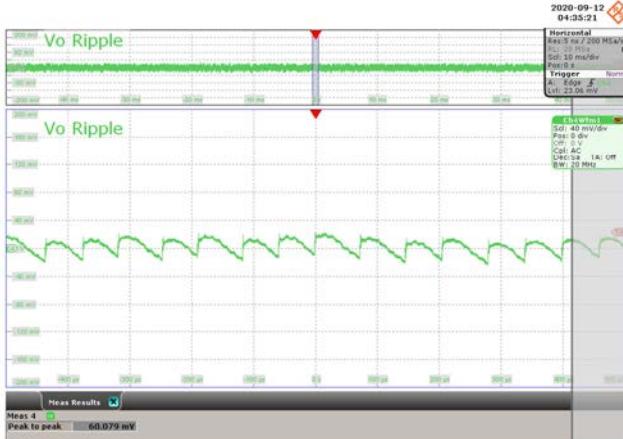
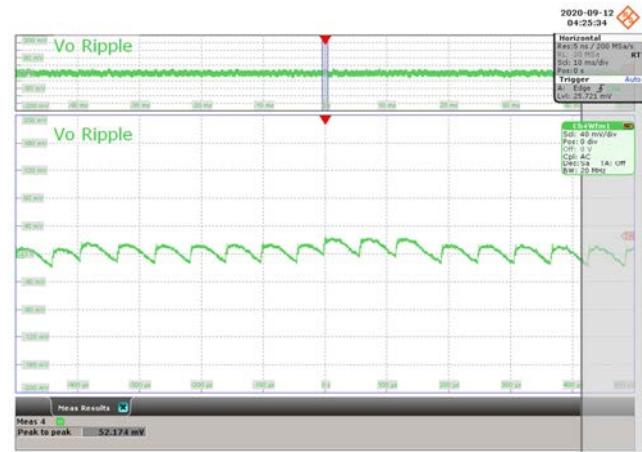
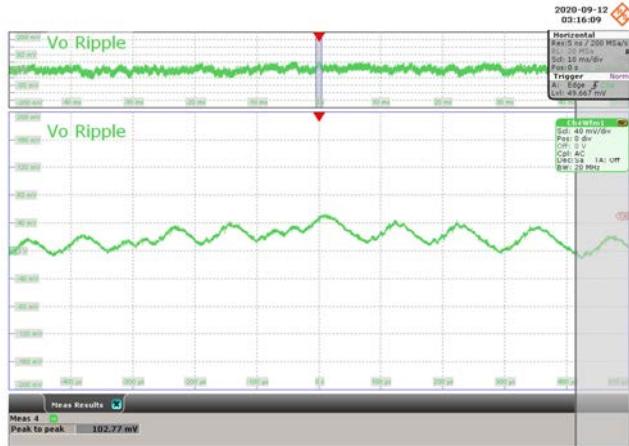


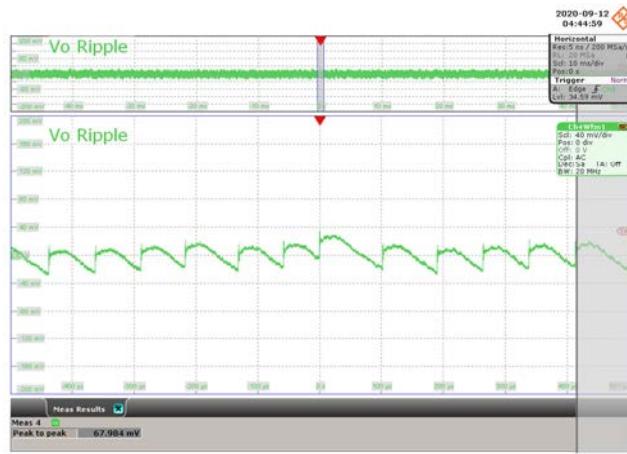
Figure 71 – Oscilloscope Probe with Probe Master (www.probemaster.com) 4987A BNC Adapter. (Modified with wires for ripple measurement, and two parallel decoupling capacitors added.)

9.5.2 Measurement Results

Note: All ripple measurements were taken at PCB end.

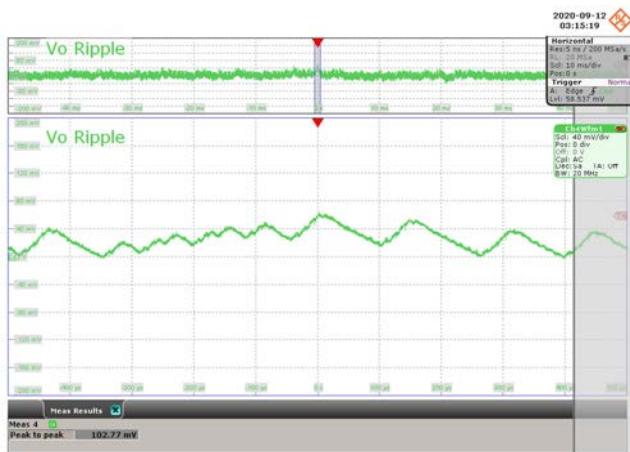
9.5.2.1 100% Load Condition



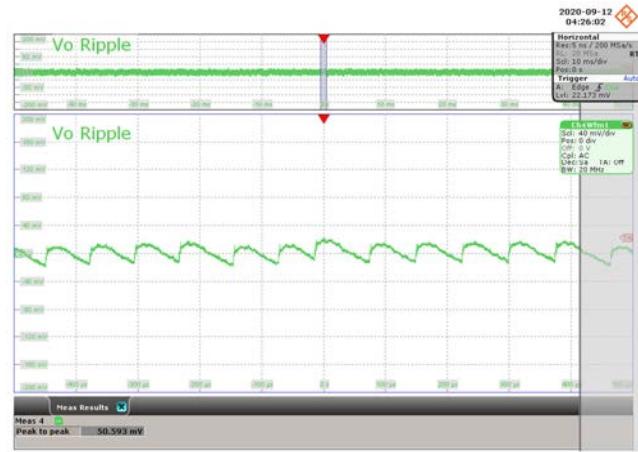
**Figure 76** – 1100 VDC Input.CH4: V_{OUT} , 40 mV / div., 10 ms / div.Zoom: 100 μ s / div.

18 V Output Ripple = 67.98 mV.

9.5.2.2 75% Load Condition

**Figure 77** – 60 VDC Input.CH4: V_{OUT} , 40 mV / div., 10 ms / div.Zoom: 100 μ s / div.

18 V Output Ripple = 102.77 mV.

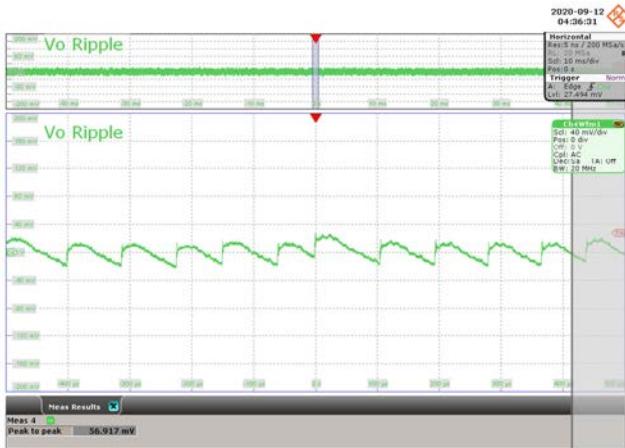


2.

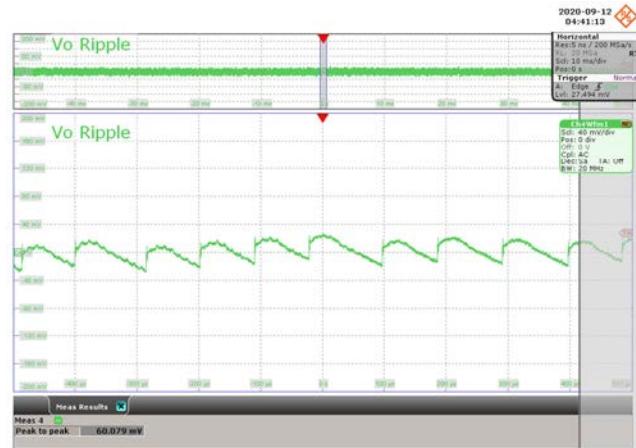
Figure 78 – 360 VDC Input.CH4: V_{OUT} , 40 mV / div., 10 ms / div.Zoom: 100 μ s / div.

18 V Output Ripple = 50.59 mV.

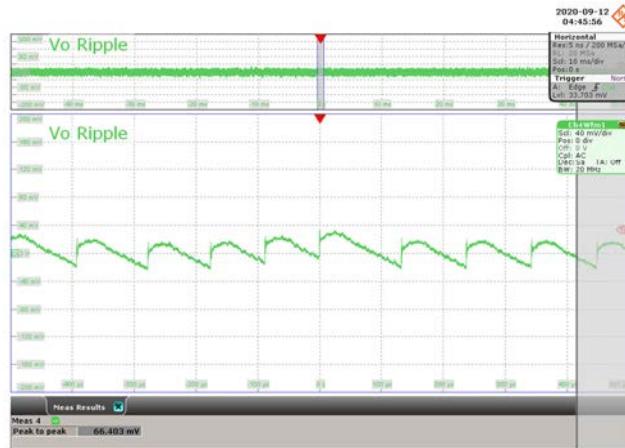


**Figure 79** – 600 VDC Input.

CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 56.91 mV.

**Figure 80** – 950 VDC Input.

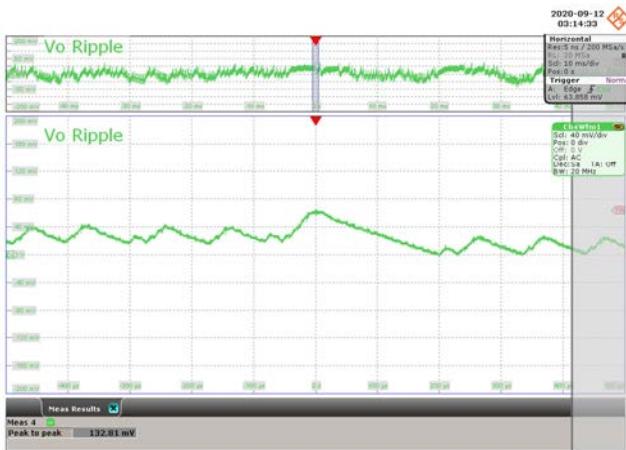
CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 60.07 mV.

**Figure 81** – 1100 VDC Input.

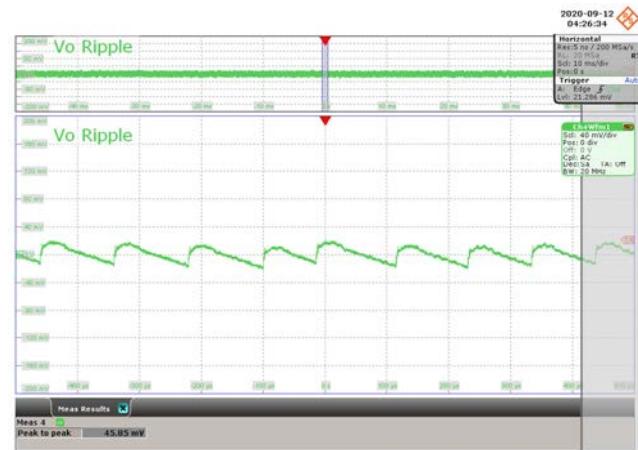
CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 66.40 mV.



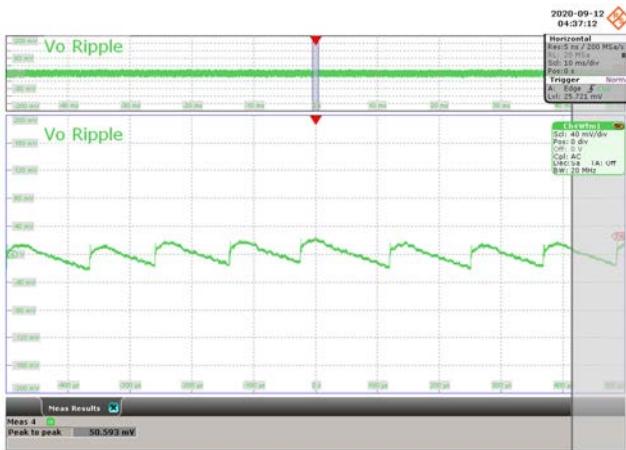
9.5.2.3 50% Load Condition

**Figure 82** – 60 VDC Input.

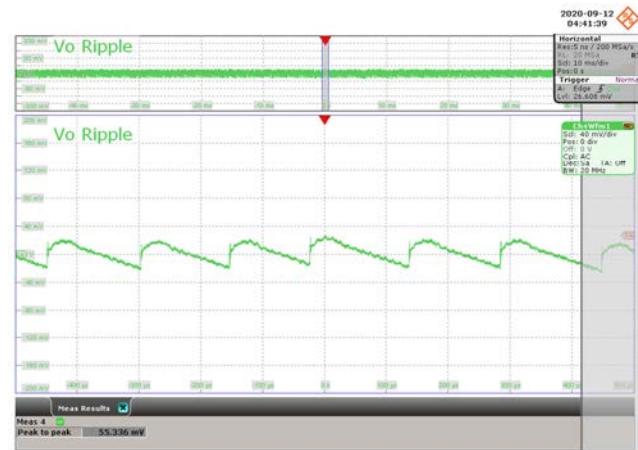
CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 132.81 mV.

**Figure 83** – 360 VDC Input.

CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 45.85 mV.

**Figure 84** – 600 VDC Input.

CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 50.59 mV.

**Figure 85** – 950 VDC Input.

CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 55.33 mV.



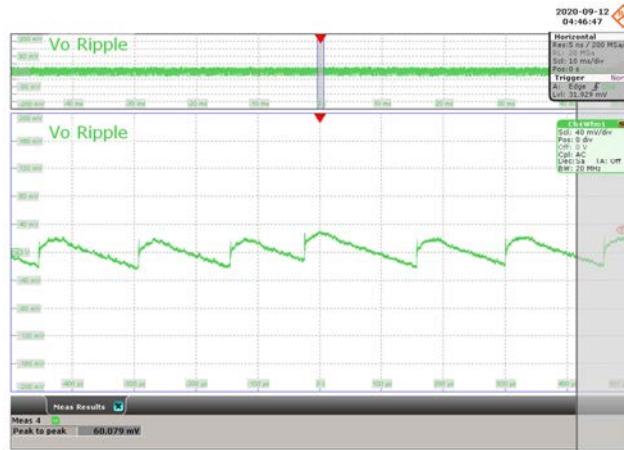


Figure 86 – 1100 VDC Input.
 CH4: V_{OUT} , 40 mV / div., 10 ms / div.
 Zoom: 100 μ s / div.
 18 V Output Ripple = 60.07 mV.

9.5.2.4 25% Load Condition

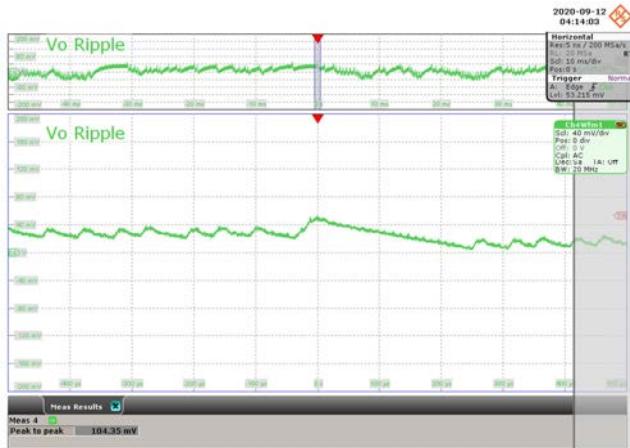


Figure 87 – 60 VDC Input.
 CH4: V_{OUT} , 40 mV / div., 10 ms / div.
 Zoom: 100 μ s / div.
 18 V Output Ripple = 104.35 mV.

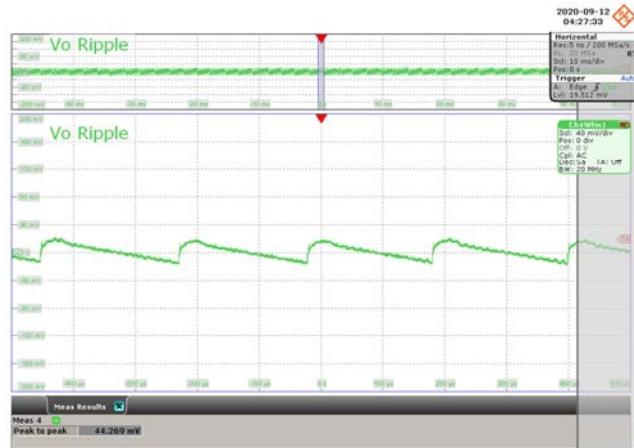
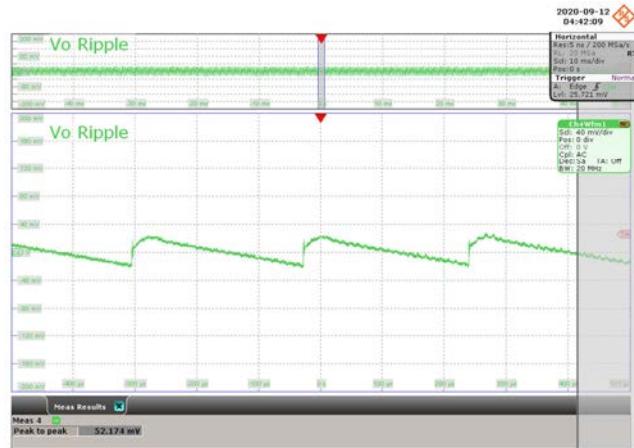


Figure 88 – 360 VDC Input.
 CH4: V_{OUT} , 40 mV / div., 10 ms / div.
 Zoom: 100 μ s / div.
 18 V Output Ripple = 44.26 mV.

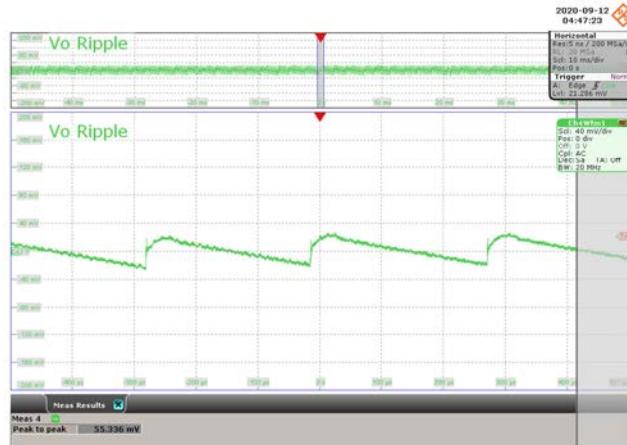


**Figure 89** – 600 VDC Input.

CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 47.43 mV.

**Figure 90** – 950 VDC Input.

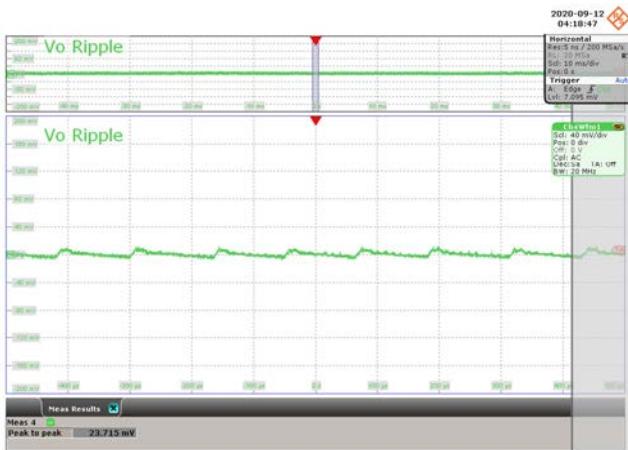
CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 52.17 mV.

**Figure 91** – 1100 VDC Input.

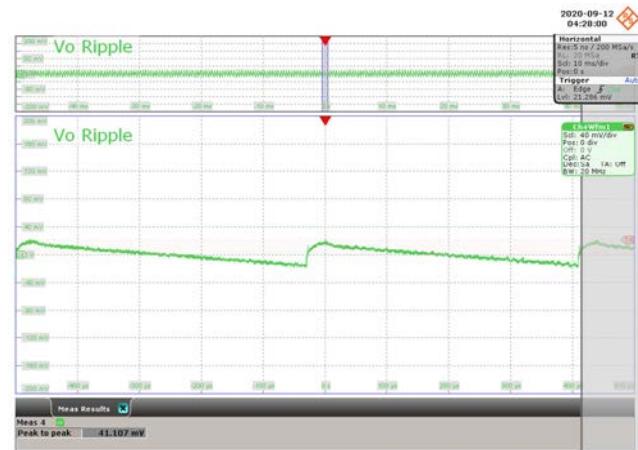
CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 55.33 mV.



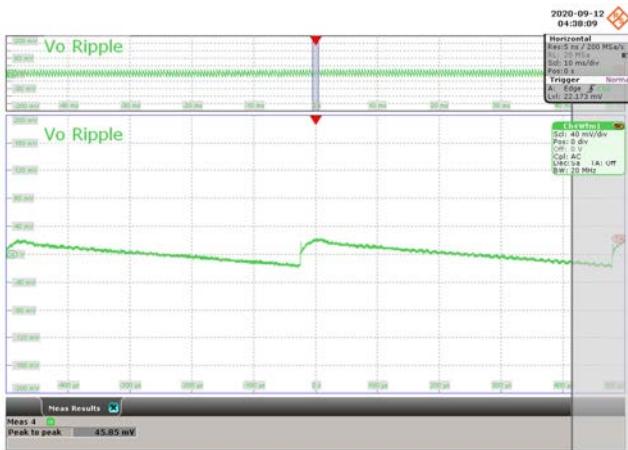
9.5.2.5 10% Load Condition

**Figure 92** – 60 VDC Input.

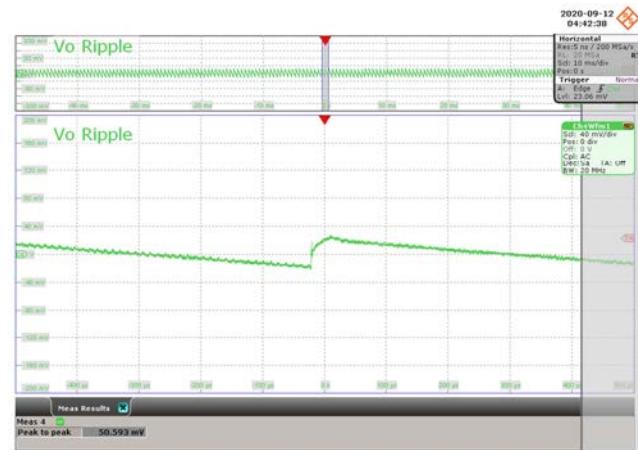
CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 23.71 mV.

**Figure 93** – 360 VDC Input.

CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 41.10 mV.

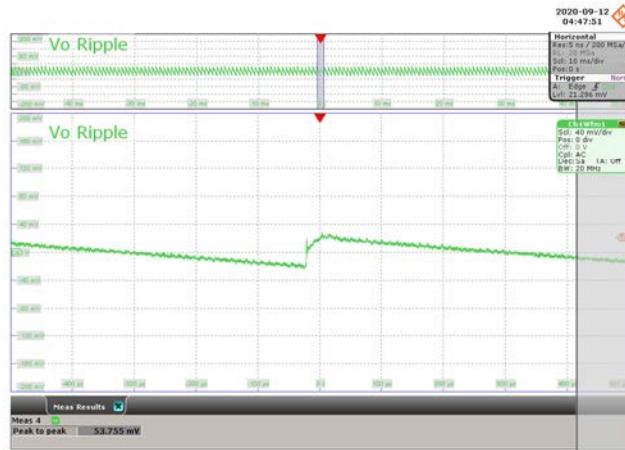
**Figure 94** – 600 VDC Input.

CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 45.85 mV.

**Figure 95** – 950 VDC Input.

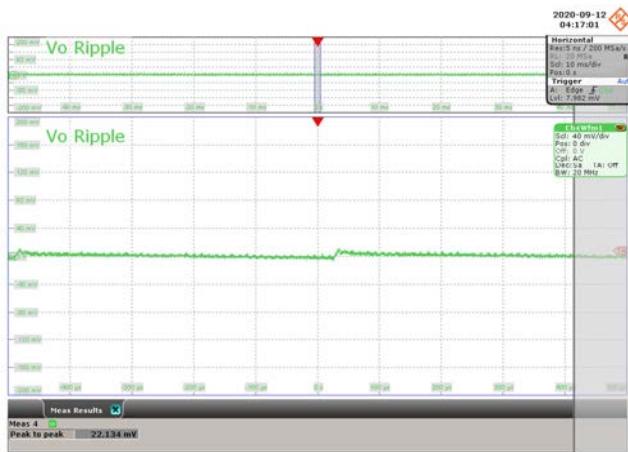
CH4: V_{OUT} , 40 mV / div., 10 ms / div.
Zoom: 100 μ s / div.
18 V Output Ripple = 50.59 mV.



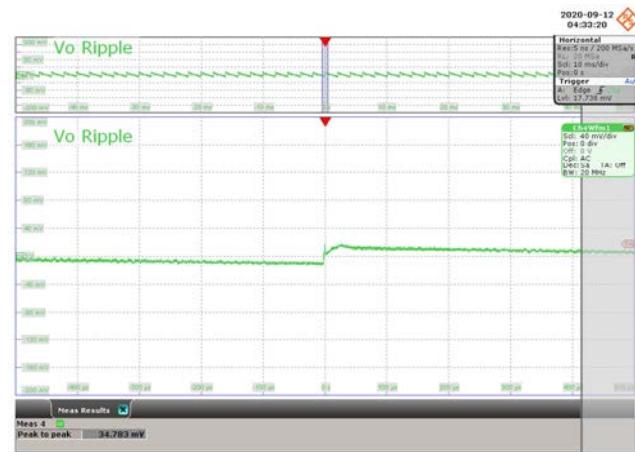
**Figure 96** – 1100 VDC Input.CH4: V_{OUT} , 40 mV / div., 10 ms / div.Zoom: 100 μ s / div.

18 V Output Ripple = 53.75 mV.

9.5.2.6 No-Load Condition

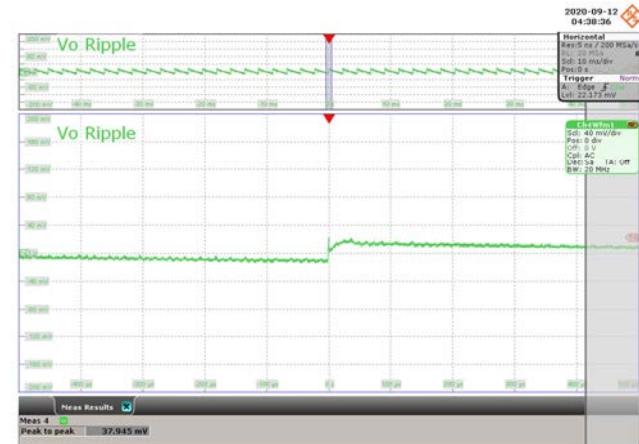
**Figure 97** – 60 VDC Input.CH4: V_{OUT} , 40 mV / div., 10 ms / div.Zoom: 100 μ s / div.

18 V Output Ripple = 22.13 mV.

**Figure 98** – 360 VDC Input.CH4: V_{OUT} , 40 mV / div., 10 ms / div.Zoom: 100 μ s / div.

18 V Output Ripple = 34.78 mV.

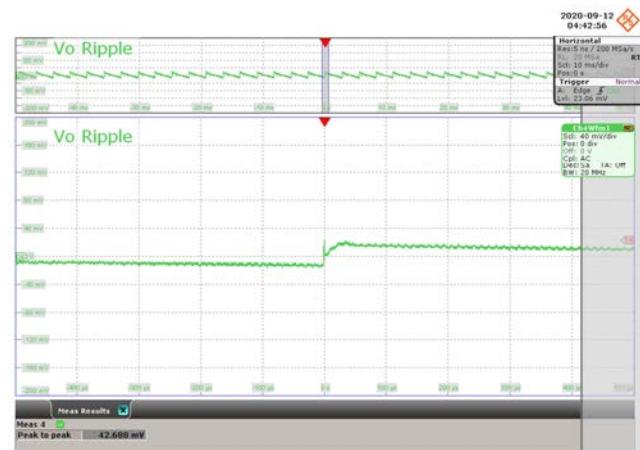




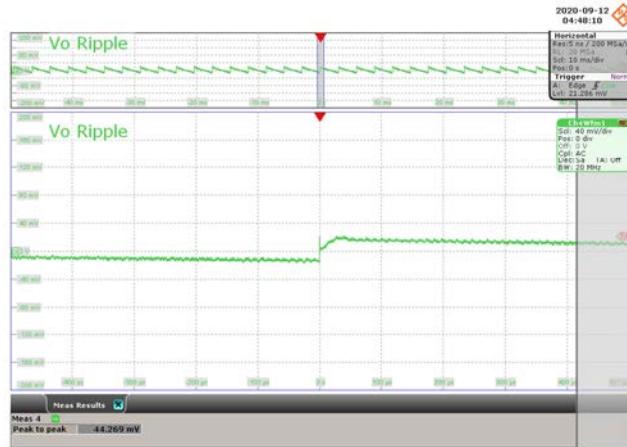
10

Figure 99 – 600 VDC Input.

CH4: $V_{o\text{ Ripple}}$, 40 mV / div., 10 ms / div.
Zoom: 100 μs / div.
18 V Output Ripple = 37.94 mV.

**Figure 100** – 950 VDC Input.

CH4: $V_{o\text{ Ripple}}$, 40 mV / div., 10 ms / div.
Zoom: 100 μs / div.
18 V Output Ripple = 42.68 mV.

**Figure 101** – 1100 VDC Input.

CH4: $V_{o\text{ Ripple}}$, 40 mV / div., 10 ms / div.
Zoom: 100 μs / div.
18 V Output Ripple = 44.269 mV.



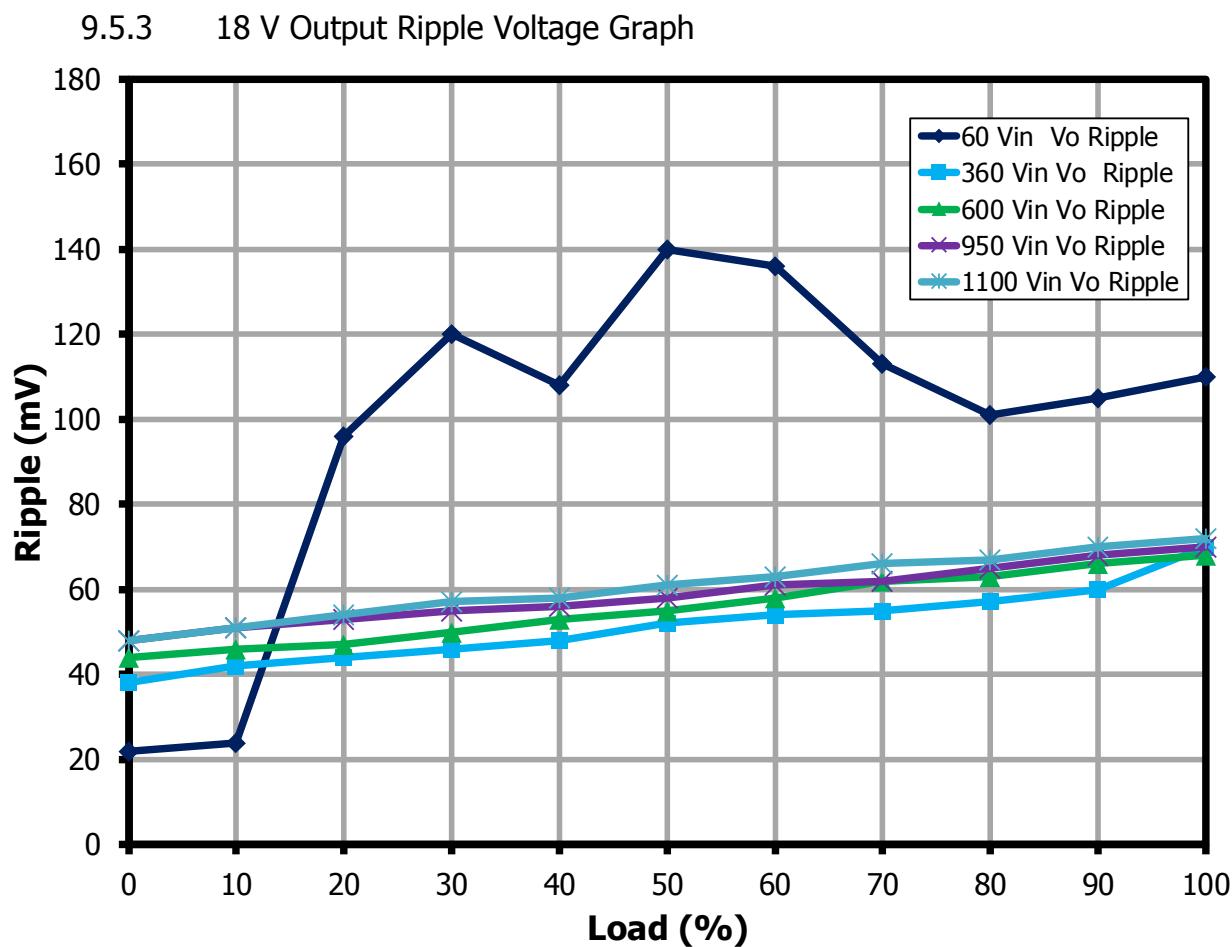


Figure 102 – 18 V Voltage Ripple (Measured at PCB End at Room Temperature).

10 Thermal Performance

10.1 *Test Set-up*

Thermal evaluation was performed under two conditions: (1) room temperature with the circuit board enclosed inside an acrylic box and (2), 105 °C ambient inside a thermal chamber. In both conditions, the circuit is soaked for one hour under full load conditions.

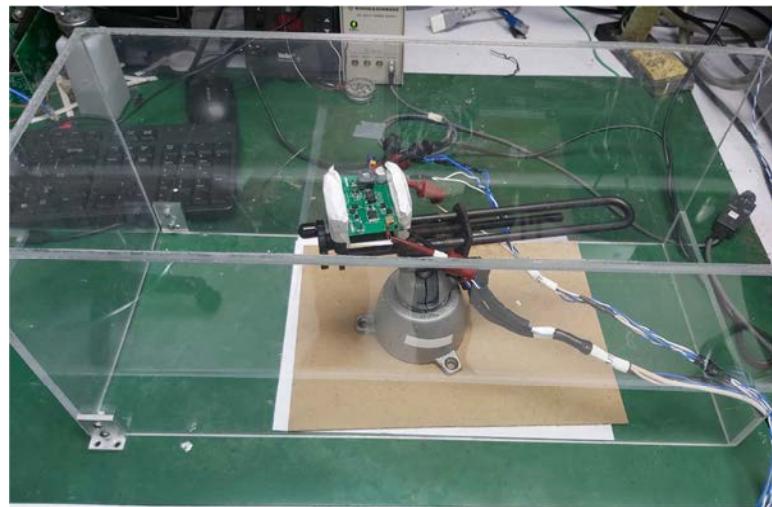


Figure 103 – Thermal Performance Set-up Using an Acrylic Box.



Figure 104 – Thermal Performance Set-up Using Thermal Chamber.

10.2 Thermal Performance at Room Temperature

10.2.1 60 VDC at Room Temperature

Test Condition: 29.7 °C ambient, soak time = 1 hour, full load

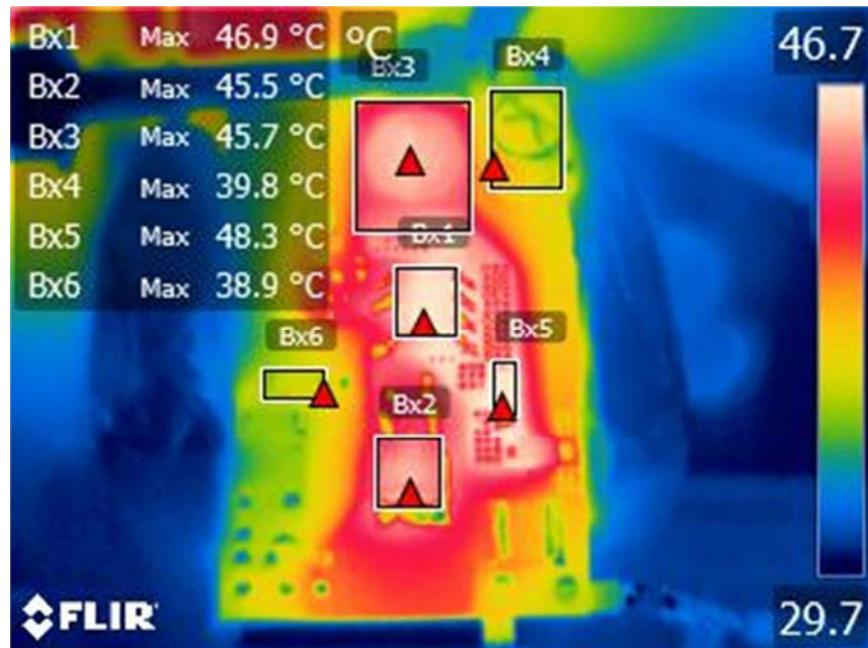


Figure 105 – Thermal Performance at 60 VDC Input.

Component	Temperature (°C)
Ambient	29.7
LNK3206GQ (U1)	46.9
StackFET (Q1)	45.5
Output Inductor (L1)	45.7
Output Capacitor (C2)	39.8
Freewheeling Diode (D1)	48.3
Clamp Zener (VR3)	38.9



10.2.2 950 VDC at Room Temperature

Test Condition: 29.3 °C ambient, soak time = 1 hour, full load

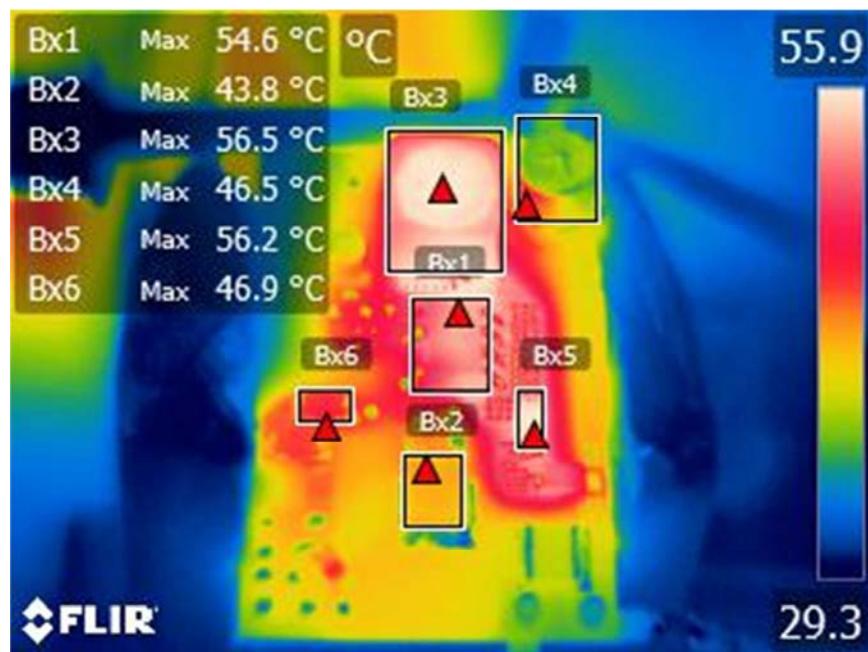


Figure 106 – Thermal Performance at 950 VDC Input.

Component	Temperature (°C)
Ambient	29.3
LNK3206GQ (U1)	54.6
StackFET (Q1)	43.8
Output Inductor (L1)	56.5
Output Capacitor (C2)	46.5
Freewheeling Diode (D1)	56.2
Clamp Zener (VR3)	46.9



10.2.3 1100 VDC at Room Temperature

Test Condition: 29.3 °C ambient, soak time = 1 hour, full load

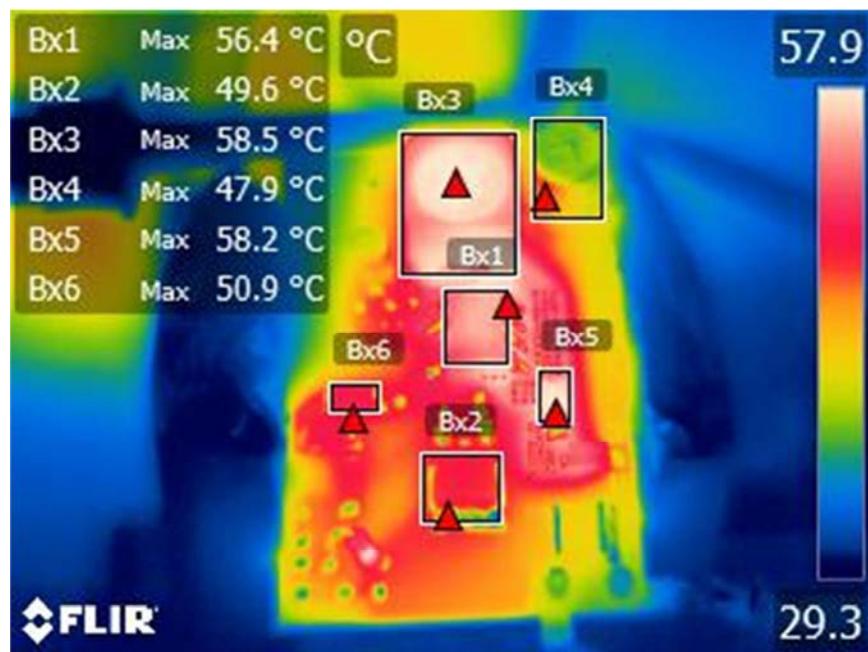


Figure 107 – Thermal Performance at 1100 VDC Input.

Component	Temperature (°C)
Ambient	29.3
LNK3206GQ (U1)	56.4
StackFET (Q1)	49.6
Output Inductor (L1)	58.5
Output Capacitor (C2)	47.9
Freewheeling Diode (D1)	58.2
Clamp Zener (VR3)	50.9



10.3 Thermal Performance at 105 °C

10.3.1 60 VDC at 105 °C Ambient

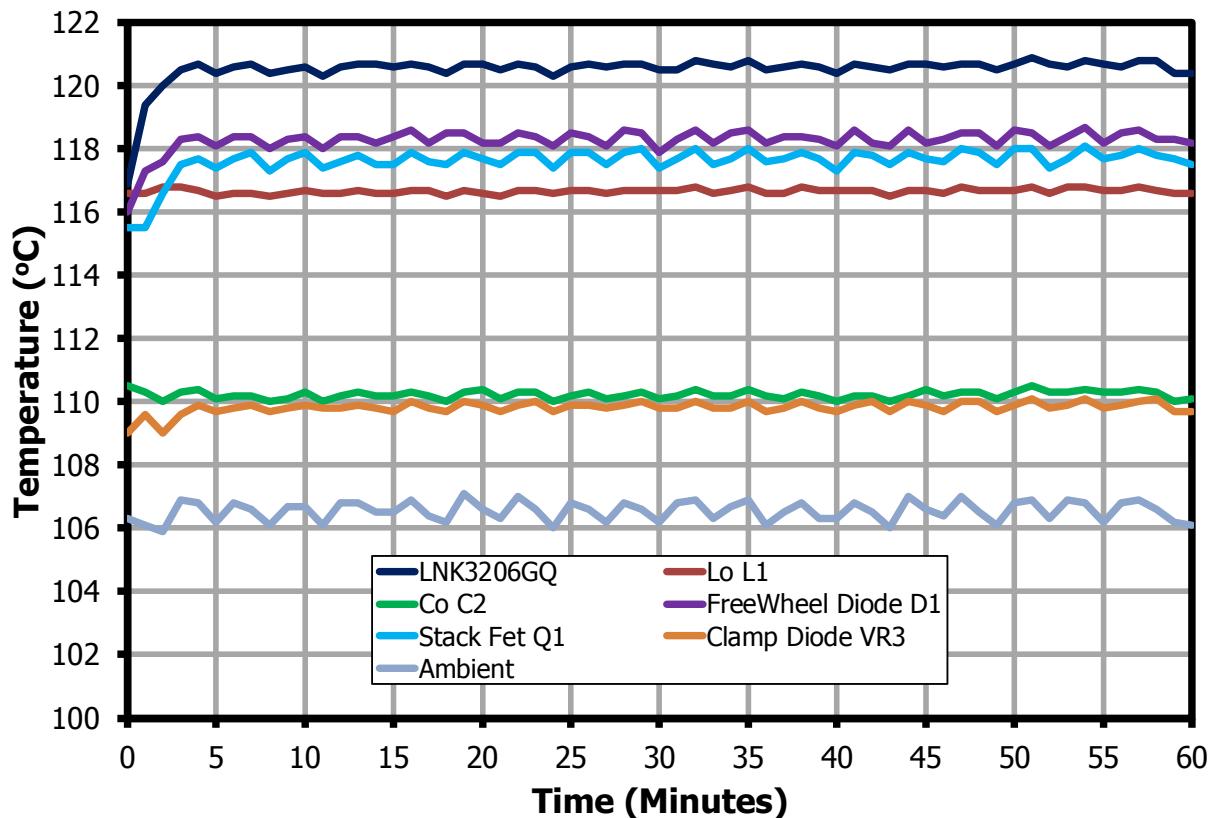


Figure 108 – Thermal Performance 60 VDC input at 105 °C.

Component	Temperature (°C)
Ambient	106.1
LNK3206GQ (U1)	120.4
StackFET (Q1)	117.5
Output Inductor (L1)	116.6
Output Capacitor (C2)	110.1
Freewheeling Diode (D1)	118.2
Clamp Zener (VR3)	109.7

10.3.2 950 VDC at 105 °C Ambient

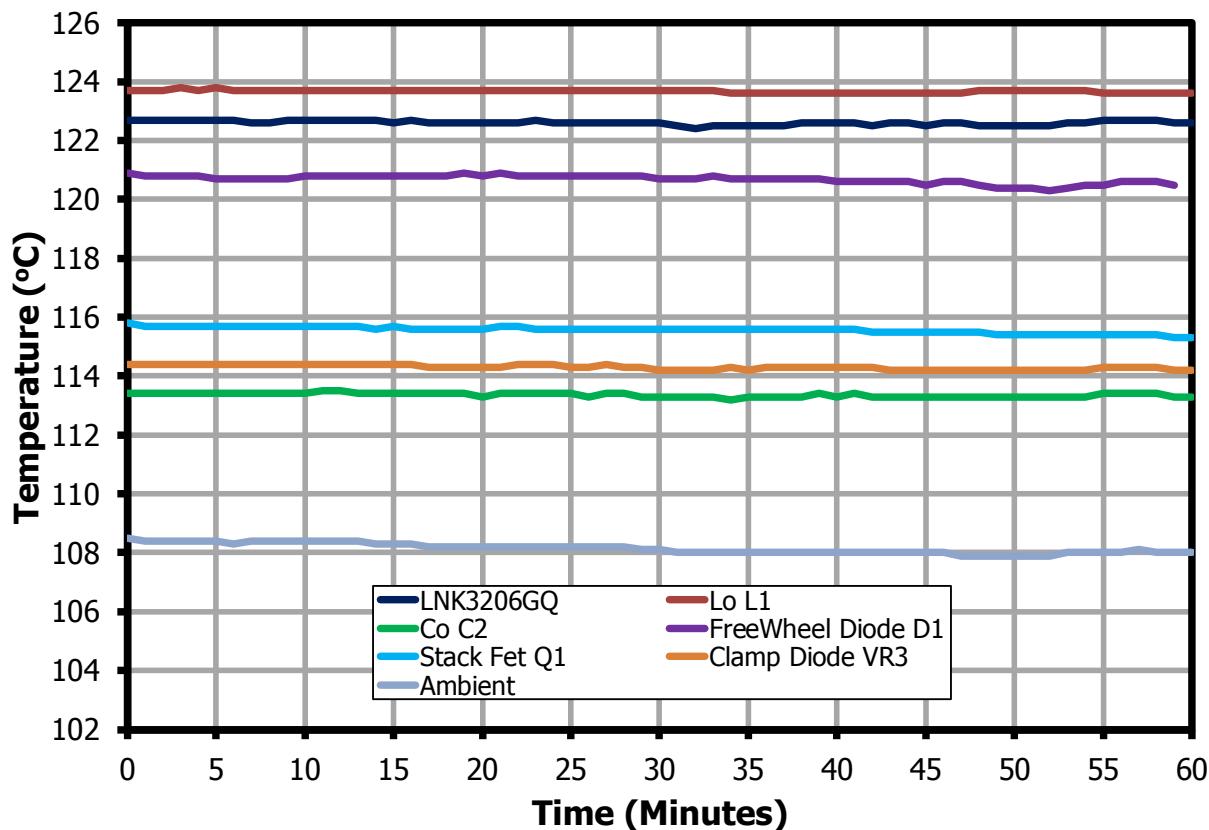


Figure 109 – Thermal Performance 950 VDC input at 105 °C.

Component	Temperature (°C)
Ambient	108
LNK3206GQ (U1)	122.6
StackFET (Q1)	115.3
Output Inductor (L1)	123.6
Output Capacitor (C2)	113.3
Freewheeling Diode (D1)	120.5
Clamp Zener (VR3)	114.2

10.3.3 1100 VDC at 105 °C Ambient

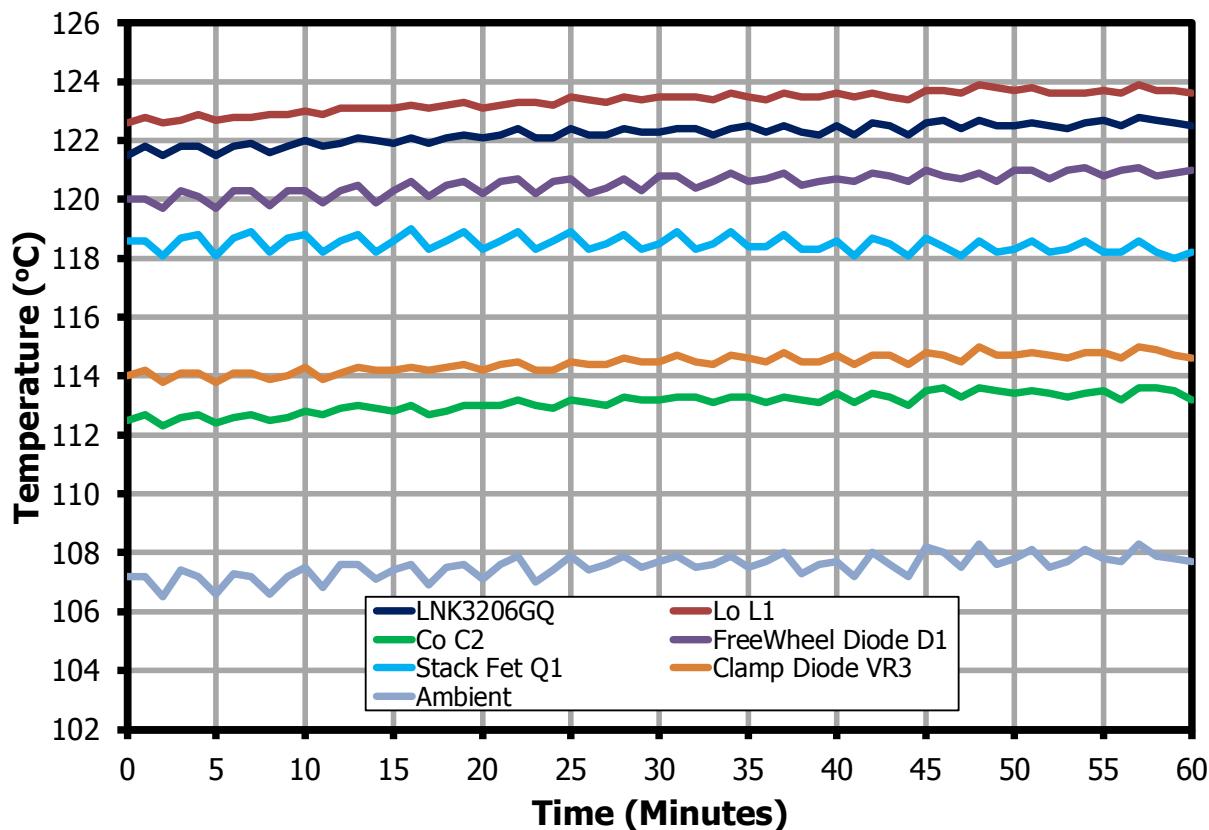


Figure 110 – Thermal Performance 1100 VDC input at 105 °C.

Component	Temperature (°C)
Ambient	107.7
LNK3206GQ (U1)	122.5
StackFET (Q1)	118.2
Output Inductor (L1)	123.6
Output Capacitor (C2)	113.2
Freewheeling Diode (D1)	121
Clamp Zener (VR3)	114.6

11 Revision History

Date	Author	Revision	Description and Changes	Reviewed
07-Jan-21	RPA	1.0	Initial Release.	Apps & Mktg
23-Jul-21	KM	1.1	Minor Formatting Change.	Mktg



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